

Module 3 : Sampling and Reconstruction

Lecture 22 : Sampling and Reconstruction of Band-Limited Signals

Objectives

Scope of this lecture:

If a Continuous Time (C.T.) signal is to be uniquely represented and recovered from its samples, then the signal must be **band-limited**. Further we have to realize that the samples must be sufficiently close and the **Sampling Rate** must bear certain relation with the highest frequency component of the original signal. In this lecture, we'll see:

- A note about Band-limited signals.
- The analyticity of time-limited and band-limited signals.
- Reconstruction of Band-limited signals - The Shannon-Whittaker-Nyquist Sampling Theorem

Band-limited signals:

A Band-limited signal is one whose Fourier Transform is non-zero on only a finite interval of the frequency axis.

Specifically, there exists a positive number **B** such that **X(f)** is non-zero only in $f \in [-B, B]$. **B** is also called the Bandwidth of the signal.

To start off, let us first make an observation about the class of Band-limited signals.

Lets consider a Band-limited signal **x(t)** having a Fourier Transform **X(f)**.

Let the interval for which **X(f)** is non-zero be $-B \leq f \leq B$.

Then,
$$x(t) = \int_{-B}^B X(f) e^{j2\pi f t} df$$
 converges.

The RHS of the above equation is differentiable with respect to **t** any number of times as the integral is performed on a bounded domain and the integrand is differentiable with respect to t. Further, in evaluating the derivative of the RHS, we can take $\frac{d}{dt}$ inside the integral.

$$\frac{dx(t)}{dt} = \int_{-B}^B (j2\pi f) X(f) e^{j2\pi f t} df$$

In general,

$$\frac{d^n x(t)}{dt^n} = \int_{-B}^B (j2\pi f)^n X(f) e^{j2\pi f t} df$$

This implies that band limited signals are **infinitely differentiable**, therefore, very **smooth**.

We now move on to see how a Band-limited signal can be reconstructed from its samples.

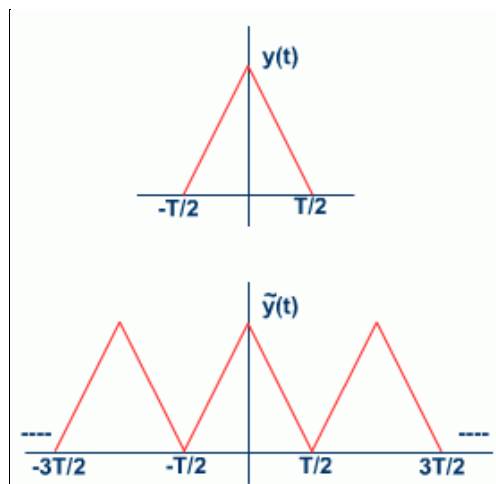
Reconstruction of Time-limited Signals

Consider first a signal **y(t)** that is **time-limited**, i.e. it is non-zero only in $[-T/2, T/2]$.

Its Fourier transform **Y(f)** is given by:

$$Y(f) = \int_{-\frac{T}{2}}^{+\frac{T}{2}} y(t) e^{-j2\pi f t} dt$$

$$= \int_{-\infty}^{+\infty} \tilde{y}(t) e^{-j2\pi f t} dt \quad \rightarrow (1)$$



Where $\tilde{y}(t)$ is the periodic extension of $y(t)$ as shown

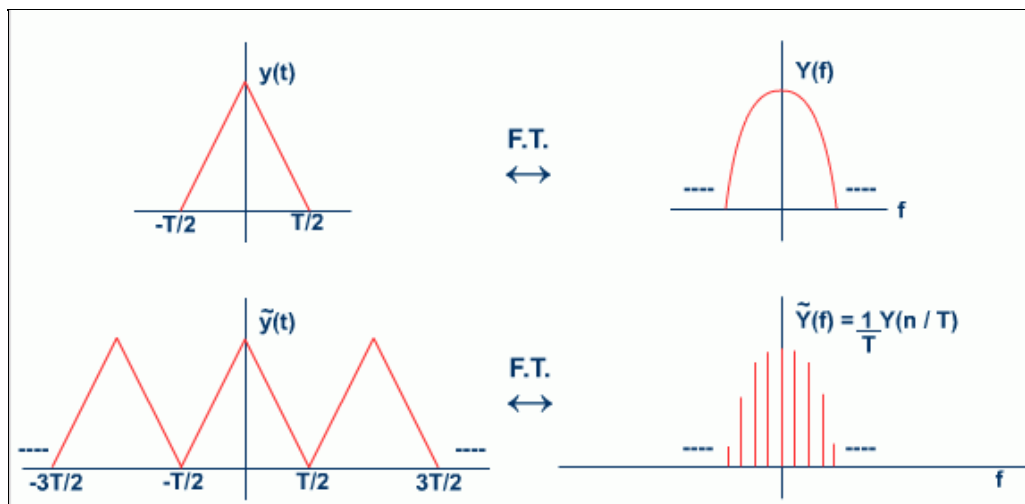
Now, Recall that the coefficients of the Fourier series for a periodic signal $x(t)$ are given by :

$$a_n = \frac{1}{T} \int_{-T/2}^{T/2} x(t) e^{-j2\pi n f_0 t} dt \quad \text{where } f_0 = \frac{1}{T} \quad \text{--- (2)}$$

Comparing (1) and (2), you will find

$$a_n = \frac{1}{T} Y\left(\frac{n}{T}\right)$$

That is, the Fourier Transform of the periodic signal $\tilde{y}(t)$ is nothing but the samples of the original transform.



Therefore, given that; $y(t)$ is time-limited in $[-T/2, T/2]$ and periodic, the entire information about $y(t)$ is contained in just **equispaced samples of its Fourier transform**! It is the dual of this result that is the basis of Sampling and Reconstruction of Band-limited signals :-

Knowing the **Fourier transform is limited** to, say $[-B, B]$, the entire information about the transform (and hence the signal) is contained in just **uniform samples of the (time) signal** !

Reconstruction of Band-limited signals

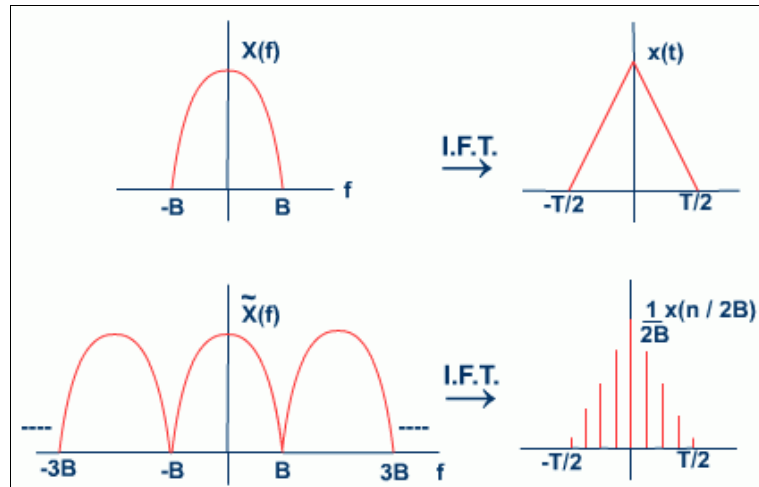
Let us now apply the dual reasoning of the previous discussion to Band-limited signals.

$x(t)$ is Band-limited, with its Fourier transform $X(f)$ being non-zero only in $[-B, B]$. The dual reasoning of the discussion in previous slide will imply that we can reconstruct $X(f)$ perfectly in $[-B, B]$ by using only the samples $x(n / 2B)$. Let's see how.

$$x(t) = \int_{-\infty}^{+\infty} X(f) e^{j2\pi f t} df = \int_{-B}^{+B} X(f) e^{j2\pi f t} df$$

$$\therefore x\left(\frac{n}{2B}\right) = \int_{-B}^B X(f) e^{j2\pi \frac{n}{2B} f} df$$

This time, $\frac{1}{2B} x\left(\frac{n}{2B}\right)$ is the $-n^{\text{th}}$ Fourier series co-efficient of $\tilde{X}(f)$, the periodic extension of $X(f)$.



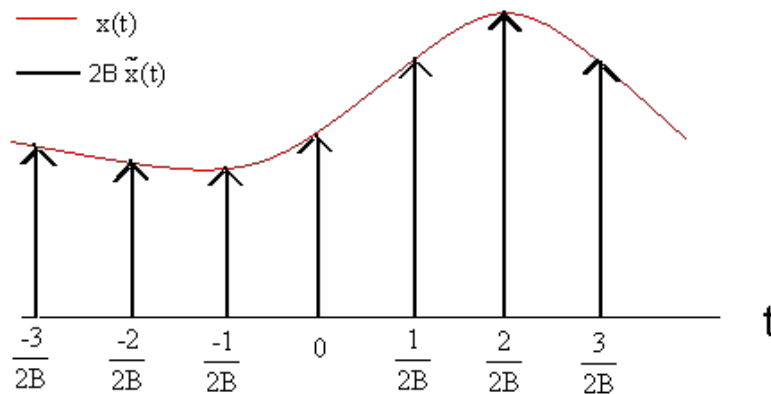
$$\therefore \tilde{X}(f) = \sum_{n=-\infty}^{\infty} \frac{1}{2B} x\left(\frac{n}{2B}\right) e^{j2\pi f \frac{n}{2B}}$$

(Fourier series in f -- fundamental period is $2B$ and $\frac{1}{2B} x\left(\frac{n}{2B}\right)$ is the $-n^{\text{th}}$ Fourier series coefficient)

What is the Fourier inverse of $\tilde{X}(f)$?

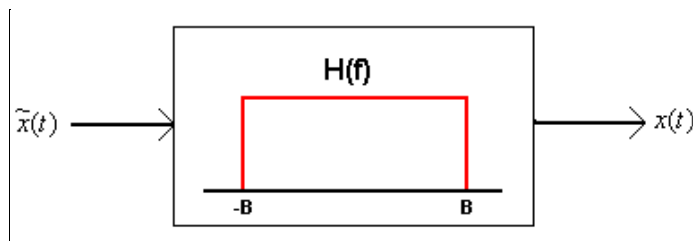
The Fourier inverse of $e^{j2\pi f t_0}$ is $\delta(t - t_0)$. Therefore, the Fourier inverse $\tilde{x}(t)$ of $\tilde{X}(f)$ is

$$\tilde{x}(t) = \sum_{n=-\infty}^{\infty} \frac{1}{2B} x\left(\frac{n}{2B}\right) \delta\left(t - \frac{n}{2B}\right)$$



Thus we see that if we multiply the original Band-limited signal with a periodic train of impulses (period $1/2B$, with impulse at the origin of strength $1/2B$) we obtain a signal whose Fourier transform is a periodic extension of the original spectrum. So how does one retrieve the original signal from $\tilde{x}(t)$? We need a mechanism that will blank out the spectrum of $\tilde{x}(t)$ in $|f| > B$, i.e: multiply the spectrum with :

$$H(f) = \begin{cases} 1 & -B \leq f \leq B \\ 0 & \text{Otherwise} \end{cases}$$



In other words, we need to feed $\tilde{x}(t)$ to an LSI system, the Fourier transform of whose impulse response is the above function (recall the convolution theorem), i.e: one whose impulse response is:

$$h(t) = \int_{-B}^B H(f) e^{j2\pi ft} df$$

An LSI system with above type of impulse response is called an Ideal Low Pass Filter .

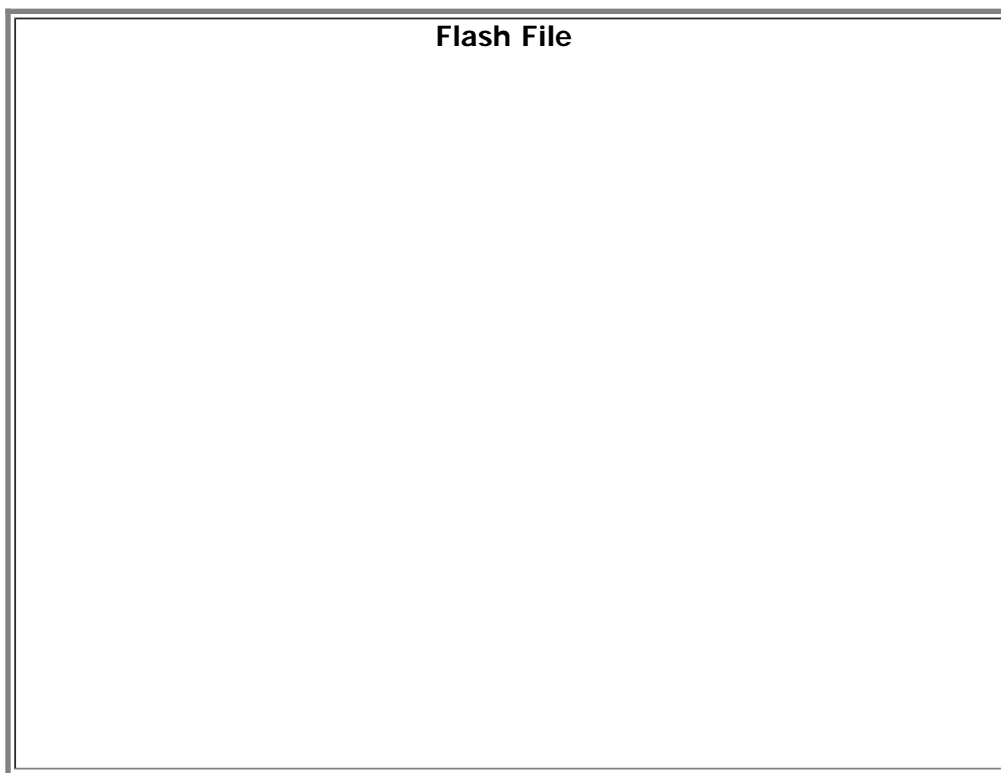
The Sampling Theorem

On the basis of our discussion so far, we may state formally the Sampling Theorem.

Shannon-Whittaker-Nyquist Sampling Theorem:

A band-limited signal with band-width B may be reconstructed perfectly from its samples, if the signal is sampled uniformly at a rate greater than $2B$.

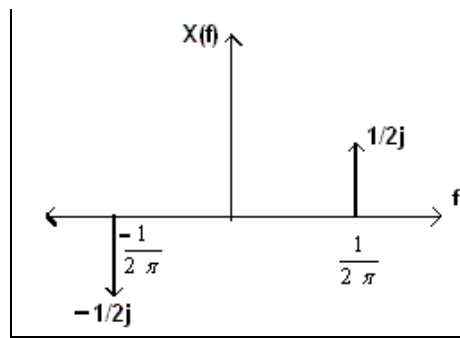
Here's an overview of the derivation of sampling theorem:



Is it essential for the sampling rate to be greater than $2B$, or is it acceptable to have a sampling rate of exactly $2B$?

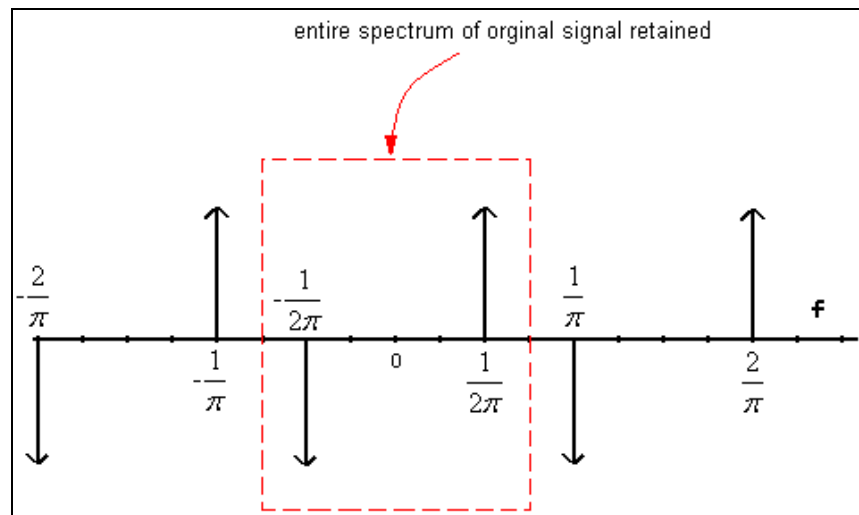
What will happen if the value of $X(f)$ at $-B$ and B are not zero? $\hat{X}(f)$ will have values at B and $-B$ different from those of $X(f)$ (due to the periodic extension). Thus the transform of the output of the ideal low pass filter will not match that of the original signal at $-B$ and B . While finite, point mismatches in the transform will not matter, problems arise if $X(f)$ has impulses at B or $-B$. Then, the output of the ideal low pass filter will be different from the original signal.

For example, consider $\sin(t)$. It has a bandwidth $\frac{1}{2\pi}$. Say we sample the signal at a rate $\frac{1}{\pi}$. What happens to all our samples? The signal has value zero at all multiples of π ! You can't possibly reconstruct the signal from these samples. What went wrong? Let's look at the Fourier Transform involved:



Note that the periodic extension (taking period to be $\frac{1}{2\pi}$) of this signal is identically zero. Thus an ideal low pass filter cannot retrieve this spectrum from its periodic extension.

This is why the Sampling theorem says one must use a sampling rate greater than $2B$, where B is the Bandwidth of the signal. Say we sample at a rate $\frac{3}{2} * \frac{1}{\pi}$. What is the Fourier transform of $\sum_{-\infty}^{\infty} (\frac{2}{3}\pi) x(\frac{2}{3}\pi n) \delta(t - \frac{2}{3}\pi n)$?



Now, an appropriate Low-pass filter can give us back the original signal !

Conclusion:

In this lecture you have learnt:

- Band-limited signals are infinitely differentiable and very smooth.
- Given that 'x(t)' is **Band-limited** with its Fourier transform 'X(f)' being non-zero only in **[-B,B]** , we can say that

$$\sum_{-\infty}^{\infty} \frac{1}{2B} x\left(\frac{n}{2B}\right) \delta\left(t - \frac{n}{2B}\right)$$

has a

spectrum that is the **periodic extension** of 'X(f)' with period $2B$.

- By passing $\sum_{-\infty}^{\infty} \frac{1}{2B} x\left(\frac{n}{2B}\right) \delta\left(t - \frac{n}{2B}\right)$ through an appropriate **Ideal Low-pass filter** one can obtain back 'x(t)'.

Shannon-Whittaker-Nyquist Sampling Theorem:

- A band-limited signal with band-width 'B' may be reconstructed perfectly from its samples, if the signal is sampled at a rate **greater** than ' $2B$ '.

Congratulations, you have finished Lecture 22.