

# Module 11: Introduction to Optimal Control

## Lecture Note 3

### 1 Linear Quadratic Regulator

Consider a linear system modeled by

$$\mathbf{x}(k+1) = A\mathbf{x}(k) + B\mathbf{u}(k), \quad \mathbf{x}(k_0) = \mathbf{x}_0$$

where  $\mathbf{x}(k) \in R^n$  and  $\mathbf{u}(k) \in R^m$ . The pair  $(A, B)$  is controllable.

The objective is to design a stabilizing linear state feedback controller  $\mathbf{u}(k) = -K\mathbf{x}(k)$  which will minimize the quadratic performance index, given by,

$$J = \sum_{k=0}^{\infty} (\mathbf{x}^T(k)Q\mathbf{x}(k) + \mathbf{u}^T(k)R\mathbf{u}(k))$$

where,  $Q = Q^T \geq 0$  and  $R = R^T > 0$ . Such a controller is denoted by  $\mathbf{u}^*$ .

We first assume that a linear state feedback optimal controller exists such that the closed loop system

$$\mathbf{x}(k+1) = (A - BK)\mathbf{x}(k)$$

is asymptotically stable.

This assumption implies that there exists a Lyapunov function  $V(\mathbf{x}(k)) = \mathbf{x}(k)^T P \mathbf{x}(k)$  for the closed loop system, for which the forward difference

$$\Delta V(\mathbf{x}(k)) = V(\mathbf{x}(k+1)) - V(\mathbf{x}(k))$$

is negative definite.

We will now use the theorem as discussed in the previous lecture which says if the controller  $\mathbf{u}^*$  is optimal, then

$$\min_{\mathbf{u}} (\Delta V(\mathbf{x}(k)) + \mathbf{x}^T(k)Q\mathbf{x}(k) + \mathbf{u}^T(k)R\mathbf{u}(k)) = 0$$

Now, finding an optimal controller implies that we have to find an appropriate Lyapunov function which is then used to construct the optimal controller.

Let us first find the  $\mathbf{u}^*$  that minimizes the function

$$f = f(\mathbf{u}(k)) = \Delta V(\mathbf{x}(k)) + \mathbf{x}^T(k)Q\mathbf{x}(k) + \mathbf{u}^T(k)R\mathbf{u}(k)$$

If we substitute  $\Delta V$  in the above expression, we get

$$\begin{aligned} f(\mathbf{u}(k)) &= \mathbf{x}^T(k+1)P\mathbf{x}(k+1) - \mathbf{x}^T(k)P\mathbf{x}(k) + \mathbf{x}^T(k)Q\mathbf{x}(k) + \mathbf{u}^T(k)R\mathbf{u}(k) \\ &= (A\mathbf{x}(k) + B\mathbf{u}(k))^T P(A\mathbf{x}(k) + B\mathbf{u}(k)) - \mathbf{x}^T(k)P\mathbf{x}(k) + \mathbf{x}^T(k)Q\mathbf{x}(k) + \mathbf{u}^T(k)R\mathbf{u}(k) \end{aligned}$$

Taking derivative of the above function with respect to  $\mathbf{u}(k)$ ,

$$\begin{aligned} \frac{\partial f(\mathbf{u}(k))}{\partial \mathbf{u}(k)} &= 2(A\mathbf{x}(k) + B\mathbf{u}(k))^T PB + 2\mathbf{u}^T R \\ &= 2\mathbf{x}^T(k)A^T PB + 2\mathbf{u}^T(k)(B^T PB + R) \\ &= \mathbf{0}^T \end{aligned}$$

The matrix  $B^T PB + R$  is positive definite since  $R$  is positive definite, thus it is invertible. Hence,

$$\mathbf{u}^*(k) = -(B^T PB + R)^{-1} B^T P A \mathbf{x}(k) = -K \mathbf{x}(k)$$

where  $K = (B^T PB + R)^{-1} B^T P A$ . Let us denote  $B^T PB + R$  by  $S$ . Thus

$$\mathbf{u}^*(k) = -S^{-1} B^T P A \mathbf{x}(k)$$

We will now check whether or not  $\mathbf{u}^*$  satisfies the second order sufficient condition for minimization. Since

$$\begin{aligned} \frac{\partial^2 f(\mathbf{u}(k))}{\partial \mathbf{u}^2(k)} &= \frac{\partial}{\partial \mathbf{u}(k)} (2\mathbf{x}^T(k)A^T PB + 2\mathbf{u}^T(k)(B^T PB + R)) \\ &= 2(B^T PB + R) > 0 \end{aligned}$$

$\mathbf{u}^*$  satisfies the second order sufficient condition to minimize  $f$ .

The optimal controller can thus be constructed if an appropriate Lyapunov matrix  $P$  is found. For that let us first find the closed loop system after introduction of the optimal controller.

$$\mathbf{x}(k+1) = (A - BS^{-1}B^T PA)\mathbf{x}(k)$$

Since the controller satisfies the hypothesis of the theorem, discussed in the previous lecture,

$$\mathbf{x}^T(k+1)P\mathbf{x}(k+1) - \mathbf{x}^T(k)P\mathbf{x}(k) + \mathbf{x}^T(k)Q\mathbf{x}(k) + \mathbf{u}^{*T}(k)R\mathbf{u}^*(k) = 0$$

Putting the expression of  $\mathbf{u}^*$  in the above equation,

$$\begin{aligned} & \mathbf{x}^T(k)(A - BS^{-1}B^T PA)^T P(A - BS^{-1}B^T PA)\mathbf{x}(k) - \mathbf{x}^T(k)P\mathbf{x}(k) + \mathbf{x}^T(k)Q\mathbf{x}(k) + \\ & \quad \mathbf{x}^T(k)A^T P BS^{-1} RS^{-1} B^T PA\mathbf{x}(k) \\ &= \mathbf{x}^T(k)A^T PA\mathbf{x}(k) - \mathbf{x}^T(k)A^T P BS^{-1} B^T PA\mathbf{x}(k) - \mathbf{x}^T(k)A^T P BS^{-1} B^T PA\mathbf{x}(k) + \\ & \quad \mathbf{x}^T(k)A^T P BS^{-1} B^T P BS^{-1} B^T PA\mathbf{x}(k) - \mathbf{x}^T(k)P\mathbf{x}(k) + \mathbf{x}^T(k)Q\mathbf{x}(k) + \\ & \quad \mathbf{x}^T(k)A^T P BS^{-1} RS^{-1} B^T PA\mathbf{x}(k) \\ &= \mathbf{x}^T(k)A^T PA\mathbf{x}(k) - \mathbf{x}^T(k)P\mathbf{x}(k) + \mathbf{x}^T(k)Q\mathbf{x}(k) - 2\mathbf{x}^T(k)A^T P BS^{-1} B^T PA\mathbf{x}(k) + \\ & \quad \mathbf{x}^T(k)A^T P BS^{-1} (B^T PB + R)S^{-1} B^T PA\mathbf{x}(k) \\ &= \mathbf{x}^T(k)A^T PA\mathbf{x}(k) - \mathbf{x}^T(k)P\mathbf{x}(k) + \mathbf{x}^T(k)Q\mathbf{x}(k) - \\ & \quad 2\mathbf{x}^T(k)A^T P BS^{-1} SS^{-1} B^T PA\mathbf{x}(k) + \mathbf{x}^T(k)A^T P BS^{-1} B^T PA\mathbf{x}(k) \\ &= \mathbf{x}^T(k)(A^T PA - P + Q - A^T P BS^{-1} B^T PA)\mathbf{x}(k) = 0 \end{aligned}$$

The above equation should hold for any value of  $\mathbf{x}(k)$ . Thus

$$A^T PA - P + Q - A^T P BS^{-1} B^T PA = 0$$

which is the well known discrete Algebraic Riccati Equation (ARE). By solving this equation we can get  $P$  to form the optimal regulator to minimize a given quadratic performance index.

**Example 1:** Consider the following linear system

$$\begin{aligned} \mathbf{x}(k+1) &= \begin{bmatrix} 0 & 1 \\ 0.5 & 0.8 \end{bmatrix} \mathbf{x}(k) + \begin{bmatrix} 0 \\ 1 \end{bmatrix} u(k), \quad \mathbf{x}_0 = \begin{bmatrix} 1 \\ 1 \end{bmatrix} \\ y(k) &= [1 \quad 0] \mathbf{x}(k) \end{aligned}$$

Design an optimal controller to minimize the following performance index.

$$J = \sum_{k=0}^{\infty} (x_1^2 + x_1 x_2 + x_2^2 + 0.1u^2)$$

Also, find the optimal cost.

*Solution:* The performance index  $J$  can be rewritten as

$$J = \sum_{k=0}^{\infty} (\mathbf{x}^T(k) \begin{bmatrix} 1 & 0.5 \\ 0.5 & 1 \end{bmatrix} \mathbf{x}(k) + 0.1u^2)$$

Thus,  $Q = \begin{bmatrix} 1 & 0.5 \\ 0.5 & 1 \end{bmatrix}$  and  $R = 0.1$ .

Let us take  $P$  as

$$P = \begin{bmatrix} p_1 & p_2 \\ p_2 & p_3 \end{bmatrix}$$

Then,

$$A^T P A - P = \begin{bmatrix} 0.25p_3 - p_1 & 0.5p_2 + 0.4p_3 - p_2 \\ 0.5p_2 + 0.4p_3 - p_2 & p_1 + 1.6p_2 + 0.64p_3 - p_3 \end{bmatrix}$$

$$A^T P A - P + Q = \begin{bmatrix} 0.25p_3 - p_1 + 1 & 0.4p_3 - 0.5p_2 + 0.5 \\ 0.4p_3 - 0.5p_2 + 0.5 & p_1 + 1.6p_2 - 0.36p_3 + 1 \end{bmatrix}$$

$$A^T P B = \begin{bmatrix} 0.5p_3 \\ p_2 + 0.8p_3 \end{bmatrix}, \quad B^T P A = [0.5p_3 \quad p_2 + 0.8p_3], \quad S = 0.1 + p_3$$

$$\begin{aligned} A^T P B S^{-1} B^T P A &= \frac{1}{0.1 + p_3} \begin{bmatrix} 0.5p_3 \\ p_2 + 0.8p_3 \end{bmatrix} [0.5p_3 \quad p_2 + 0.8p_3] \\ &= \frac{1}{0.1 + p_3} \begin{bmatrix} 0.25p_3^2 & 0.5p_2p_3 + 0.4p_3^2 \\ 0.5p_2p_3 + 0.4p_3^2 & p_2^2 + 1.6p_2p_3 + 0.64p_3^2 \end{bmatrix} \end{aligned}$$

The discrete ARE is

$$A^T P A - P + Q - A^T P B S^{-1} B^T P A = 0$$

Or,

$$\begin{bmatrix} 0.25p_3 - p_1 + 1 - \frac{0.25p_3^2}{0.1+p_3} & 0.4p_3 - 0.5p_2 + 0.5 - \frac{0.5p_2p_3 + 0.4p_3^2}{0.1+p_3} \\ 0.4p_3 - 0.5p_2 + 0.5 - \frac{0.5p_2p_3 + 0.4p_3^2}{0.1+p_3} & p_1 + 1.6p_2 - 0.36p_3 + 1 - \frac{p_2^2 + 1.6p_2p_3 + 0.64p_3^2}{0.1+p_3} \end{bmatrix} = \begin{bmatrix} 0 & 0 \\ 0 & 0 \end{bmatrix}$$

We can get three equations from the discrete ARE. These are

$$0.25p_3 - p_1 + 1 - \frac{0.25p_3^2}{0.1 + p_3} = 0$$

$$0.4p_3 - 0.5p_2 + 0.5 - \frac{0.5p_2p_3 + 0.4p_3^2}{0.1 + p_3} = 0$$

$$p_1 + 1.6p_2 - 0.36p_3 + 1 - \frac{p_2^2 + 1.6p_2p_3 + 0.64p_3^2}{0.1 + p_3} = 0$$

Since the above three equations comprises three unknown parameters, these parameters can be solved uniquely, as

$$p_1 = 1.0238, \quad p_2 = 0.5513, \quad p_3 = 1.9811$$

The optimal control law can be found out as

$$\begin{aligned} u^*(k) &= -(R + B^T P B)^{-1} B^T P A \mathbf{x}(k) \\ &= -[0.4760 \quad 1.0265] \mathbf{x}(k) \\ &= -0.4760x_1(k) - 1.0265x_2(k) \end{aligned}$$

The optimal cost can be found as

$$\begin{aligned} J &= \mathbf{x}_0^T P \mathbf{x}_0 \\ &= \begin{bmatrix} 1 & 1 \end{bmatrix} \begin{bmatrix} 1.0238 & 0.5513 \\ 0.5513 & 1.9811 \end{bmatrix} \begin{bmatrix} 1 \\ 1 \end{bmatrix} \\ &= 4.1075 \end{aligned}$$