

## Lecture 22 :

<http://www.ee.iitm.ac.in/~nagendra/cadinfo.html>

-  $0.18\mu m$  CMOS process models

Exercise:

$$\left\{ \frac{1.8\mu m}{0.18\mu m}, \frac{3.6\mu m}{0.36\mu m}, \frac{7.2\mu m}{0.72\mu m}, \frac{14.4\mu m}{1.44\mu m} \right\}$$

\* Plot  $I_D$  vs.  $V_{GS}$  of nMOS & pMOS transistors  
(in saturation region)

\* Plot  $I_D$  vs.  $V_{DS}$  of nMOS & pMOS transistors  
(for a range of  $V_{GS}$  values -  $V_{GS}$ : parameter)

Channel length is very short :-

$$I_D = \frac{\mu C_{ox}}{2} \frac{W}{L} (V_{GS} - V_T)^2$$

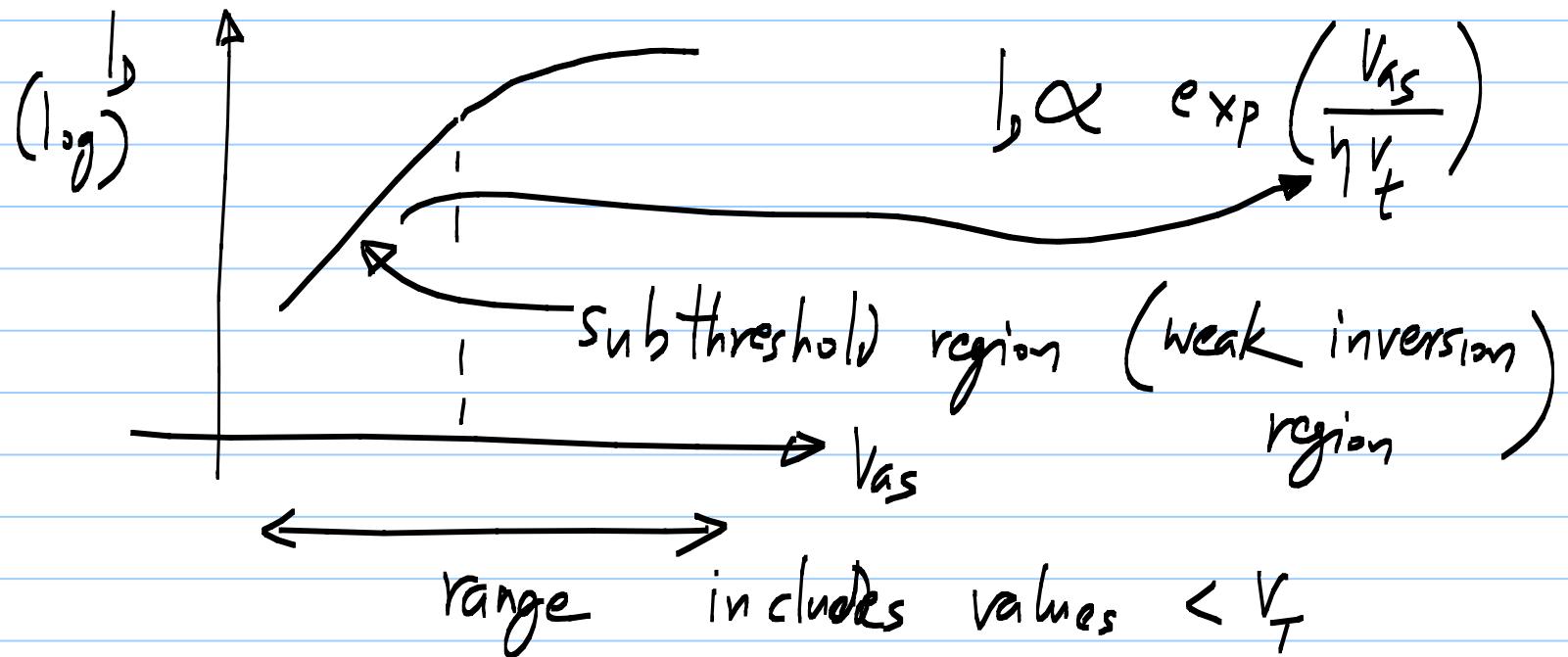
$$\frac{\mu (V_{GS} - V_T)}{L} = v$$

$$= W C_{ox} (V_{GS} - V_T) v_{SAT}$$

saturation velocity

## Current in sub threshold region.

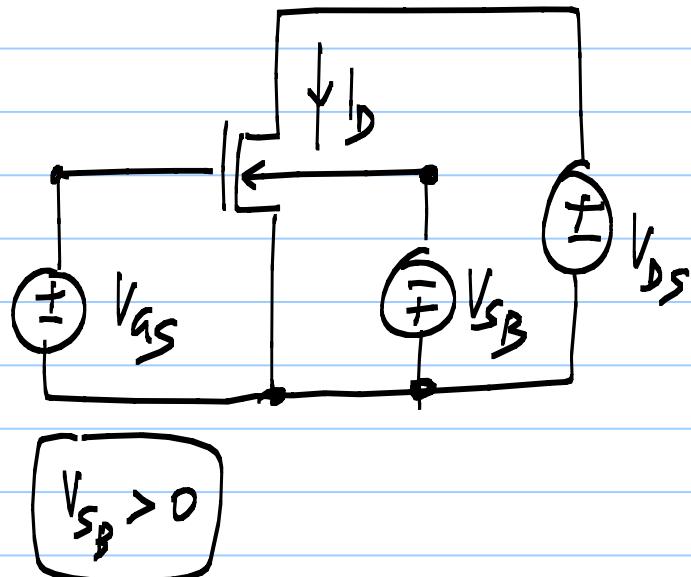
\* Plot  $I_s$  vs.  $V_{ds}$  (on a log scale).



\* Find the effect of  $V_{SB}$

\* Plot  $I_D$  vs.  $V_{SB}$

( $V_{DS}$ ,  $I_{GS}$ : strong inversion & saturation)

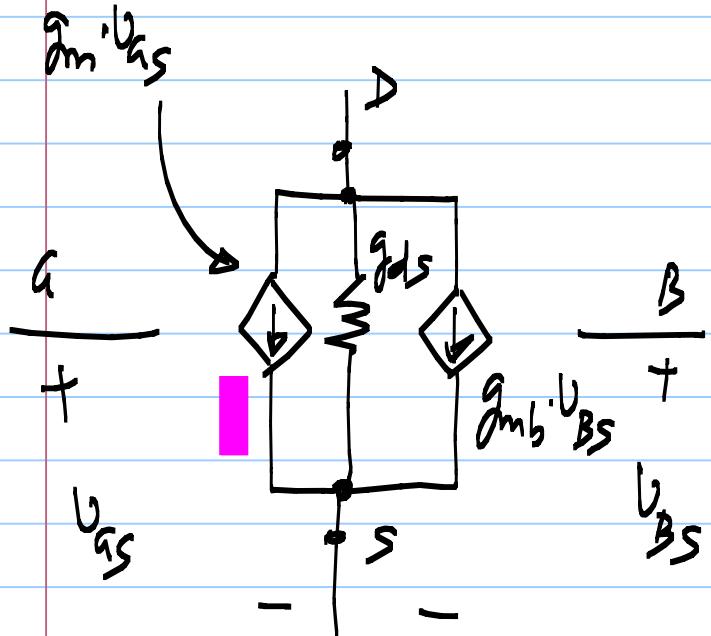


\* Plot  $I_D$  vs.  $V_{DS}$  for

different values of  $V_{SB}$

( $V_{SB}$ : parameter)

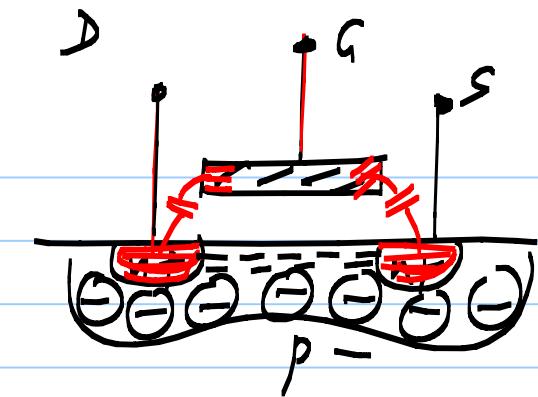
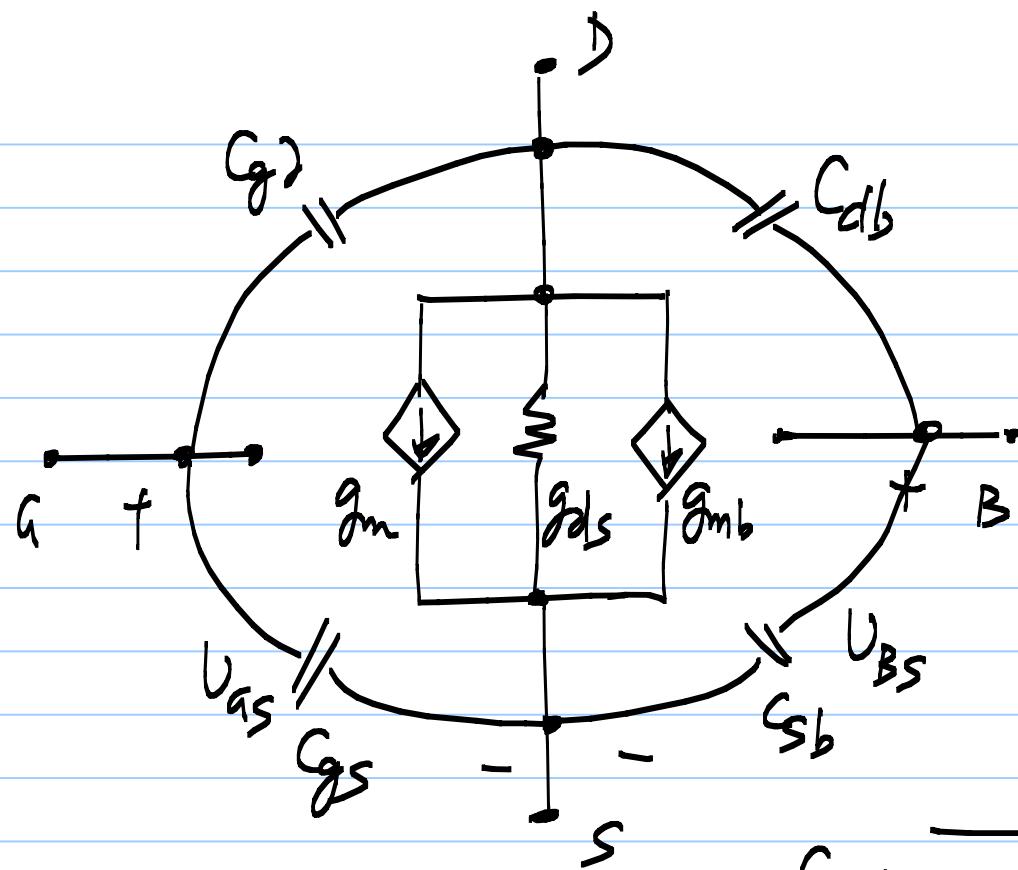
## dc Small signal model of the MOS transistor:



$$I_D = f(V_{GS}, V_{DS}, V_{BS})$$

$$\Delta I_D = \frac{\partial I_D}{\partial V_{GS}} \Big|_{op} U_{GS} + \frac{\partial I_D}{\partial V_{DS}} \Big|_{op} U_{DS} + \frac{\partial I_D}{\partial V_{BS}} \Big|_{op} U_{BS}$$

The equation shows the small signal change in drain current  $\Delta I_D$  as the sum of three terms. Each term is the partial derivative of the drain current with respect to a bias voltage, evaluated at operating point (op), multiplied by the small signal change in that voltage. Red arrows point from the labels  $g_m$ ,  $g_{mb}$ , and  $g_{ds}$  to the respective partial derivatives  $\frac{\partial I_D}{\partial V_{GS}}$ ,  $\frac{\partial I_D}{\partial V_{DS}}$ , and  $\frac{\partial I_D}{\partial V_{BS}}$ .



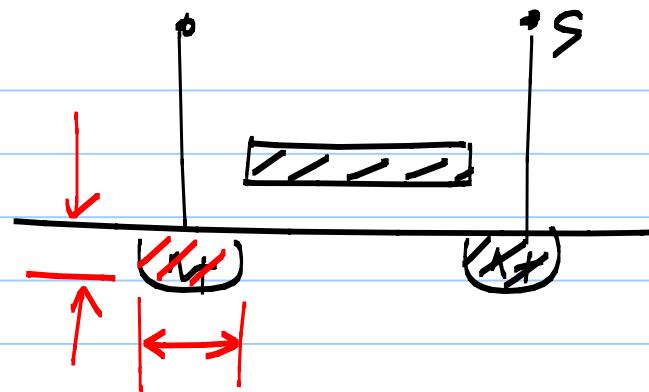
$C_{gs}, C_{sb}$ : intrinsic  
capacitances

Triode	Saturation
$C_{gd} : \frac{1}{2} \cdot W \cdot L \cdot C_{ox}$	$0 + C'_v \cdot W$
$C_{gs} : \frac{1}{2} W L C_{ox}$	$\frac{2}{3} W L C_{ox} + C'_v \cdot W$

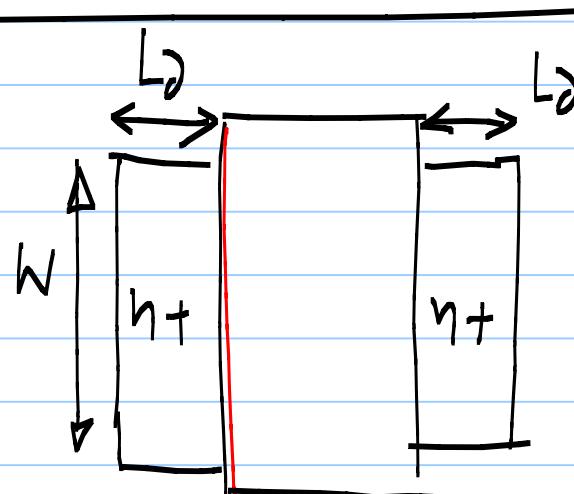
$C_{sb}, C_{db} :$

$$\underbrace{\hspace{10em}}$$

Area capacitance



+  
perimeter capacitance



$$L_d = 2 \cdot L_{min}$$

$$C_{sb} = W \cdot L_d \cdot C_{area}$$

(A<sub>s</sub>)

$$2(W + L_d) \cdot C_{perim}$$

(P<sub>s</sub>)

$$C_{db} = A_d \cdot C_{area} + P_d \cdot C_{peri}$$

Small signal model:

$$g_m = \frac{2I_D}{V_{AS}-V_T} \quad (\text{strong inv.}) ; \quad \frac{I_D}{\gamma V_T} \quad (\text{weak inv.})$$

dc {

$$g_{DS} = \lambda \cdot I_D$$
$$g_{m_b} = g_m \cdot \frac{\partial V_T}{\partial V_{SB}}$$

Capacitances {

$$C_{gd} = 0 + C_{ov'} \cdot W$$
$$C_{gs} = \frac{2}{3} C_{ox} WL + C_{ov'} \cdot W$$
$$C_{fb} = A_s \cdot C_{j,area} + P_s \cdot C_{j,peri}$$
$$C_{db} = A_d \cdot C_{j,area} + P_d \cdot C_{j,peri}$$

$$I_D = \frac{\mu_n C_{ox}}{2} \cdot \frac{W}{L} (V_{GS} - V_T)^2 (1 + \lambda V_{DS})$$

$$\begin{aligned}
 g_m &= \frac{\partial I_D}{\partial V_{GS}} = \mu_n C_{ox} \cdot \frac{W}{L} (V_{GS} - V_T) (1 + \lambda V_{DS}) \\
 &\approx \mu_n C_{ox} \frac{W}{L} (V_{GS} - V_T) \\
 &= \frac{2 \cdot I_D}{V_{GS} - V_T} \\
 &= \sqrt{2 \cdot \mu_n C_{ox} \frac{W}{L} \cdot I_D}
 \end{aligned}$$

$$g_m = \frac{2I_D}{V_{GS} - V_T}$$

Square law model  
is valid.

$$V_{GS} - V_T > 6 \cdot V_T$$

(thermal voltage)

$$\approx 150mV$$

Subthreshold region

$$I_D = I_0 \cdot \exp\left(\frac{V_{GS}}{\eta \cdot V_T}\right)$$

$$\frac{\partial I_D}{\partial V_{GS}} = \frac{I_D}{\eta \cdot V_T}$$

$$g_{DS} = \frac{\partial I_D}{\partial V_{DS}} = \underbrace{\frac{M C_0 X}{2} \frac{W}{L} (V_{GS} - V_T)^2}_{\lambda} \cdot \lambda$$

$$= \underline{\underline{\lambda I_D}}$$

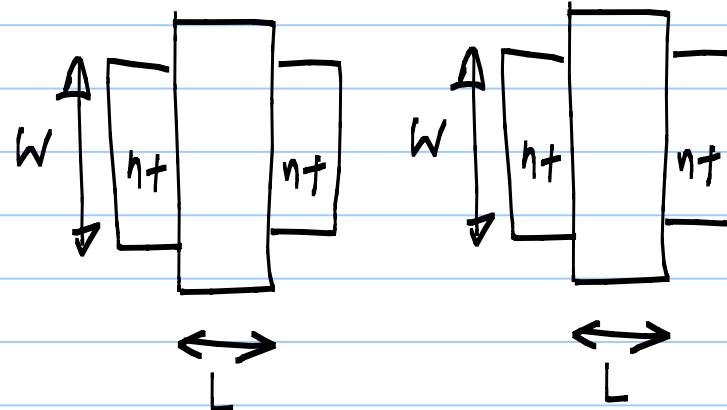
$$g_{mB} = \frac{\partial I_D}{\partial V_{BS}} = \frac{\partial I_D}{\partial V_T} \cdot \frac{\partial V_T}{\partial V_{BS}} = + M C_0 X \frac{W}{L} (V_{GS} - V_T) (1 + \lambda V_{BS})$$

$$= \underline{\underline{g_m \left( \frac{\partial V_T}{\partial V_{SB}} \right)}}$$

$$\frac{\partial V_T}{\partial V_{SB}} = \frac{1}{2} \dots \frac{1}{4}$$

MOS transistor mismatch;

For a given  $V_{GS}$ ,  $V_{DS}$ ,  $V_{SB}$



Same current

$$I = \left\{ \frac{M_Cox}{2} \frac{W}{L} \right\} (V_{GS} - V_T)^2$$

$$= \frac{\beta}{2} (V_{GS} - V_T)^2$$

$V_T$  mismatch :  $\sigma_{V_T} = \sigma (V_{T_1} - V_{T_2}) = \frac{A V_T}{\sqrt{WL}}$

$\beta$  mismatch :  $\sigma_{\beta} = \sigma \left( \frac{\beta_1 - \beta_2}{\beta_0} \right) = \frac{A \beta}{\sqrt{WL}}$