

# An introduction to Information Theory

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## Lecture #8: Coding of sources with memory



## Outline of the lecture

- Discrete stationery source (DSS)



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- Discrete stationery source (DSS)
- Block to variable length coding of DSS



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- Discrete stationery source (DSS)
- Block to variable length coding of DSS
- Coding of positive integers



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- Discrete stationery source (DSS)
- Block to variable length coding of DSS
- Coding of positive integers
- Elias-Willems Source Coding



## Discrete stationery source

- A K-ary discrete stationary source (DSS) is a device that emits a sequence  $U_i, i = 1, 2, \dots$  of K-ary random variables such that for every  $n \geq 1$  and every  $L \geq 1$ , the random vectors  $[U_1, U_2, \dots, U_L]$  and  $[U_{n+1}, U_{n+2}, \dots, U_{n+L}]$  have the same probability distribution.



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- A DSS is said to have finite memory  $\mu$  if, for all  $n > \mu$ ,

$$P_{U_n|U_1, \dots, U_{n-1}}(u_n|u_1, \dots, u_{n-1}) = P_{U_n|U_{n-\mu}, \dots, U_{n-1}}(u_n|u_{n-\mu}, \dots, u_{n-1})$$

holds for all choices of  $u_1, \dots, u_n$  and if  $\mu$  is the smallest nonnegative integer such that this holds



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- The discrete memory source (DMS) is a special case of DSS with memory  $\mu = 0$ .



## Discrete stationery source

- For DSS with memory  $\mu$ ,

$$H(U_n|U_1, \dots, U_{n-1}) = H(U_n|U_{n-\mu}, \dots, U_{n-1})$$

for all  $n > \mu$ .



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$$H_L(U) = \frac{1}{L} H(U_1, U_2, \dots, U_L)$$



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- Another way for defining uncertainty per letter of output sequence is  $H(U_L|U_1 U_2 \dots U_{L-1})$



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- (iv)  $\lim_{L \rightarrow \infty} H(U_L|U_1 \cdots U_{L-1}) = \lim_{L \rightarrow \infty} H_L(U) = H_\infty(U)$



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Proof:

- (ii)  $H(U_{L+1}|U_1 \cdots U_L) \leq H(U_{L+1}|U_2 \cdots U_L) = H(U_L|U_1 \cdots U_{L-1})$



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- (ii)  $H(U_{L+1}|U_1 \dots U_L) \leq H(U_{L+1}|U_2 \dots U_L) = H(U_L|U_1 \dots U_{L-1})$
- (i)

$$\begin{aligned} H(U_1 U_2 \dots U_L) &= H(U_1) + H(U_2|U_1) + \dots + H(U_L|U_1 \dots U_{L-1}) \\ &\geq L H(U_L|U_1 \dots U_{L-1}) \\ \implies H_L(U) &\geq H(U_L|U_1 \dots U_{L-1}) \end{aligned}$$



## Discrete stationary source

Proof:

- (iii)

$$\begin{aligned} H(U_1 U_2 \dots U_{L+1}) &= H(U_1 \dots U_L) + H(U_{L+1}|U_1 \dots U_L) \\ &\leq H(U_1 \dots U_L) + H(U_L|U_1 \dots U_{L-1}) \\ &\leq H(U_1 \dots U_L) + H_L(U) \\ &= L H_L(U) + H_L(U) \\ \implies \frac{H(U_1 U_2 \dots U_{L+1})}{L+1} &\leq H_L(U) \\ H_{L+1}(U) &\leq H_L(U) \end{aligned}$$



## Discrete stationary source

Proof:

- (iii)

$$\begin{aligned}H(U_1 U_2 \cdots U_{L+1}) &= H(U_1 \cdots U_L) + H(U_{L+1} | U_1 \cdots U_L) \\ &\leq H(U_1 \cdots U_L) + H(U_L | U_1 \cdots U_{L-1}) \\ &\leq H(U_1 \cdots U_L) + H_L(U) \\ &= L H_L(U) + H_L(U) \\ \Rightarrow \frac{H(U_1 U_2 \cdots U_{L+1})}{L+1} &\leq H_L(U) \\ H_{L+1}(U) &\leq H_L(U)\end{aligned}$$

- (iv) Since  $H_L(U) \geq 0$  and  $H(U_L | U_1 \cdots U_{L-1}) \geq 0$  and a non-increasing sequence bounded below always has a limit. Thus (ii) and (iii) ensure that both  $\lim_{L \rightarrow \infty} H(U_L | U_1 \cdots U_{L-1})$  and  $\lim_{L \rightarrow \infty} H_L(U)$  always exist. (i) implies that

$$\lim_{L \rightarrow \infty} H(U_L | U_1 \cdots U_{L-1}) \leq \lim_{L \rightarrow \infty} H_L(U) = H_\infty(U)$$

## Discrete stationary source

Proof (contd.):

- Now we will have to show that equality holds. We know that

$$\begin{aligned}H(U_1 \cdots U_L \cdots U_{L+n}) &= H(U_1 \cdots U_L) + H(U_{L+1} | U_1 \cdots U_L) + \cdots \\ &\quad + H(U_{L+n} | U_1 \cdots U_{L+n-1}) \\ &\leq H(U_1 \cdots U_L) + n H(U_{L+1} | U_1 \cdots U_L) \\ \frac{H(U_1 \cdots U_L \cdots U_{L+n})}{n+L} &\leq \frac{H(U_1 \cdots U_L)}{n+L} + \frac{n}{n+L} H(U_{L+1} | U_1 \cdots U_L)\end{aligned}$$

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- Taking the limit as  $n \rightarrow \infty$  on both sides we get,

$$H_\infty(U) \leq H(U_{L+1}|U_1 \cdots U_L)$$



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- Taking the limit as  $n \rightarrow \infty$  on both sides we get,

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- Now taking the limit as  $L \rightarrow \infty$  on both sides we get,

$$H_\infty(U) \leq \lim_{L \rightarrow \infty} H(U_{L+1}|U_1 \cdots U_L)$$



## Block to variable length coding of DSS



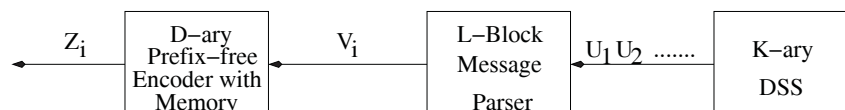
- If block length  $L$  messages from a DSS are encoded by a  $D$ -ary prefix-free code, where the code used for each message may depend on previous messages, then the length  $W_i$  of the codeword for the  $i$ -th message satisfies

$$\frac{E[W_i]}{L} \geq \frac{H_\infty(U)}{\log D}$$

where  $H_\infty(U)$  is the uncertainty per letter of the DSS.



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Proof:

- Since we are using  $L$ -block message parser, the  $i$ -th message is encoded as

$$V_i = [U_{iL-L+1}, \dots, U_{iL-1}, U_{iL}]$$



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- Given that  $[V_1, \dots, V_{i-1}] = [v_1, \dots, v_{i-1}]$ , it follows that the length  $W_i$  of the codeword  $Z_i$  for the message  $V_i$  must satisfy

$$E[W_i | [V_1, \dots, V_{i-1}] = [v_1, \dots, v_{i-1}]] \geq \frac{H(V_i | [V_1, \dots, V_{i-1}] = [v_1, \dots, v_{i-1}])}{\log D}$$



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- Multiplying by  $P_{V_1 \dots V_{i-1}}(v_1, \dots, v_{i-1})$  and summing over all values  $v_1, \dots, v_{i-1}$  gives

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- We can write

$$\begin{aligned} H(V_i | V_1 \dots V_{i-1}) &= H(U_{iL-L+1} \dots U_{iL} | U_1 \dots U_{iL-L}) \\ &= H(U_{iL-L+1} | U_1 \dots U_{iL-L}) + \dots + \\ &\quad H(U_{iL} | U_1 \dots U_{iL-1}) \\ &\geq LH(U_{iL} | U_1 \dots U_{iL-1}) \\ &\geq LH_\infty(U) \end{aligned}$$



## Coding of positive integers

- Binary representation

n	B(n)
1	1
2	1 0
3	1 1
4	1 0 0
5	1 0 1
6	1 1 0
7	1 1 1
8	1 0 0 0



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- Every codeword as 1 as the first digit.
- The length of the codeword  $L(n) = \lfloor \log_2 n \rfloor + 1$ .



## Coding of positive integers

- Elias code 1

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- No of leading zeros tell how many digits follow the leading 1.



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- This is a prefix-free code.
- No of leading zeros tell how many digits follow the leading 1.
- The length of the codeword  $L_1(n) = 2\lfloor \log_2 n \rfloor + 1$ .



# Coding of positive integers

- Elias code 2

n	$B_2(n)$
1	1
2	0 1 0 0
3	0 1 0 1
4	0 1 1 0 0
5	0 1 1 0 1
6	0 1 1 1 0
7	0 1 1 1 1
8	0 0 1 0 0 0 0 0



## Coding of positive integers

- Elias code 2

$n$	$B_2(n)$
1	1
2	0 1 0 0
3	0 1 0 1
4	0 1 1 0 0
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- Elias code 2 begins with  $B_1(L(n))$  (Elias code 1 to signal the number of following digits in the codeword) followed by  $B(n)$  with leading 1 removed.



## Coding of positive integers

- Elias code 2



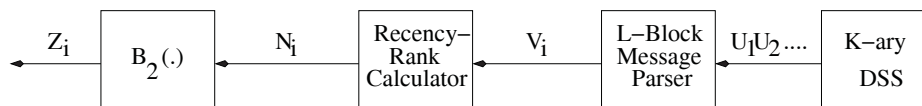
# Coding of positive integers

- Elias code 2
- The length of the codeword  $L_2(n)$  is given by

$$\begin{aligned}L_2(n) &= L_1(L(n)) + L(n) - 1 \\ &= L_1(\lfloor \log_2 n \rfloor + 1) + \lfloor \log_2 n \rfloor \\ &= \lfloor \log_2 n \rfloor + 2\lfloor \log_2(\lfloor \log_2 n \rfloor + 1) \rfloor + 1\end{aligned}$$



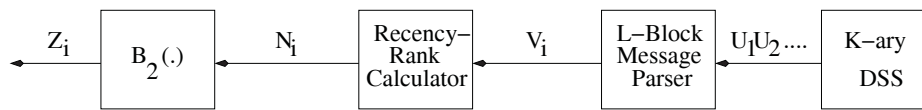
# Elias-Willems Source Coding



- The message seen most recently has recency rank of 1, next different message value most recently seen has recency rank 2 and so on.



# Elias-Willems Source Coding

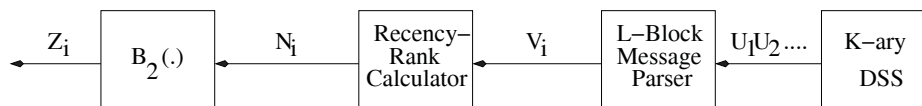


- The message seen most recently has recency rank of 1, next different message value most recently seen has recency rank 2 and so on.
- Consider  $L = 2$ , and  $K = 2$ , i.e. four possible message values are 00, 01, 10, 11. Consider the following sequence of past messages

$$\cdots | V_{i-2} | V_{i-1} = \cdots | 11 | 01 | 10 | 01 | 01 | 00 | 01$$



# Elias-Willems Source Coding



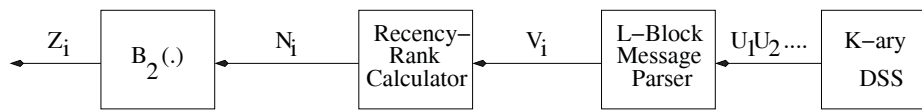
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- In this example, message 01 has a recency rank of 1, message 00 has a recency rank of 2, message 10 has a recency rank of 3, and message 11 has a recency rank of 4.



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- In this example, message 01 has a recency rank of 1, message 00 has a recency rank of 2, message 10 has a recency rank of 3, and message 11 has a recency rank of 4.
- In Elias Willems source coding scheme, recency rank calculator assigns a recency rank,  $N_i$  to the message  $V_i$ .



# Elias-Willems Source Coding

- For the Elias-Willems binary coding scheme for block length  $L$  messages from a  $K$ -ary ergodic discrete stationary source, the codeword length  $W_i$  for all  $i$  sufficiently large satisfies

$$\frac{E[W_i]}{L} \leq H_L(U) + \frac{2}{L} \log_2(LH_L(U) + 1) + \frac{1}{L}$$



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Proof:

- We define time  $\Delta_i$  to the recent occurrence of message  $V_i$  as  $\Delta_i = \delta$ , where if  $V_i = v$ ,  $V_{i-\delta}$  is the most recent past message with the same value of  $v$ . Also  $N_i \leq \Delta_i$ .



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- We can write

$$E[\Delta_i | V_i = v] = \frac{1}{P_V(v)}$$





## Elias-Willems Source Coding

- Codeword length  $W_i$  can be upper bounded by

$$\begin{aligned}W_i = L_2(N_i) &\leq L_2(\Delta_i) \\ &\leq \log_2(\Delta_i) + 2 \log_2(\log_2(\Delta_i) + 1) + 1\end{aligned}$$



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- Taking the conditional expectation, given that  $V_i = v$  and applying Jensen's inequality, we get

$$E[W_i|V_i = v] \leq \log_2(E[\Delta_i|V_i = v]) + 2 \log_2(\log_2(E[\Delta_i|V_i = v]) + 1) + 1$$



## Elias-Willems Source Coding

- Codeword length  $W_i$  can be upper bounded by

$$\begin{aligned}W_i = L_2(N_i) &\leq L_2(\Delta_i) \\ &\leq \log_2(\Delta_i) + 2 \log_2(\log_2(\Delta_i) + 1) + 1\end{aligned}$$

- Taking the conditional expectation, given that  $V_i = v$  and applying Jensen's inequality, we get

$$E[W_i | V_i = v] \leq \log_2(E[\Delta_i | V_i = v]) + 2 \log_2(\log_2(E[\Delta_i | V_i = v]) + 1) + 1$$

- Substituting the value of  $E[\Delta_i | V_i = v]$  in the above expression, we get

$$E[W_i | V_i = v] \leq -\log_2(P_V(v)) + 2 \log_2(-\log_2(P_V(v)) + 1) + 1$$



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- Multiplying the above expression by  $P_V(v)$ , summing over  $v$  and using Jensen's inequality, we get

$$E[W_i] \leq H(V) + 2 \log_2(H(V) + 1) + 1$$



## Elias-Willems Source Coding

- Also

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- Substituting this value of  $H(V)$  in the previous expression and dividing by  $L$ , we get

$$\frac{E[W_i]}{L} \leq H_L(U) + \frac{2}{L} \log_2(LH_L(U) + 1) + \frac{1}{L}$$

