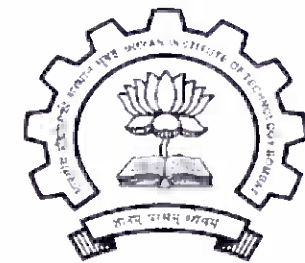
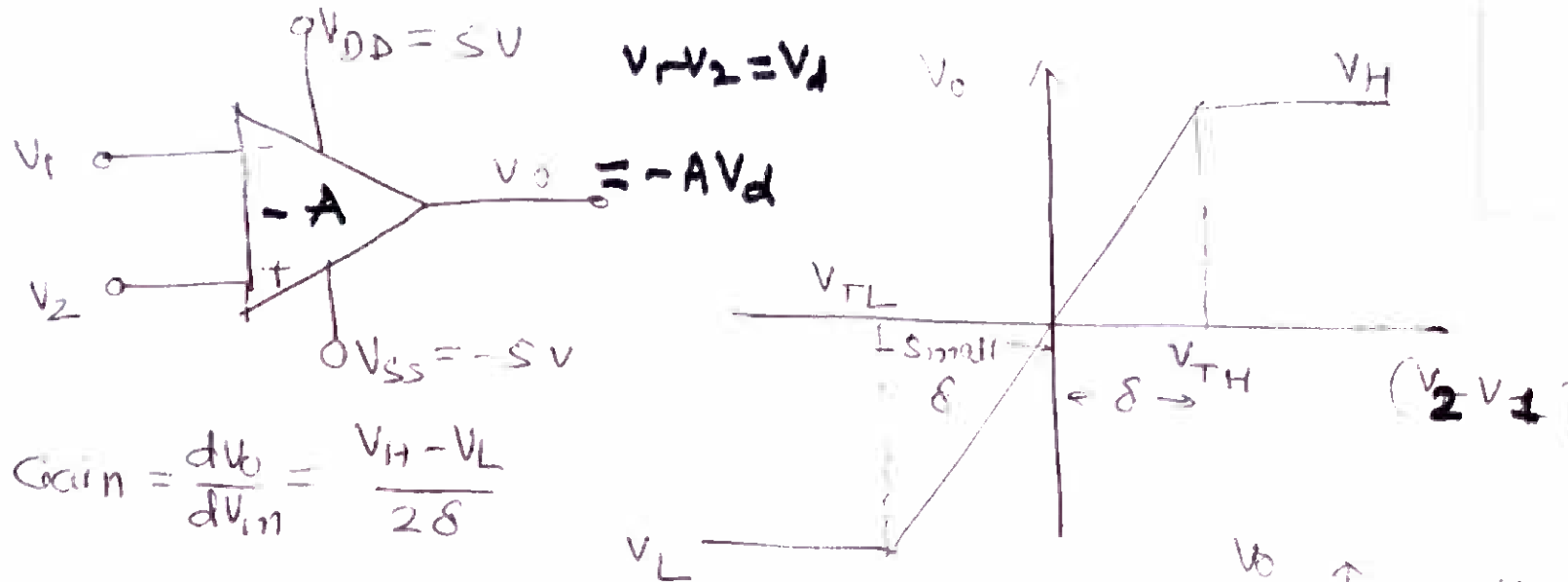


COMPARATORS

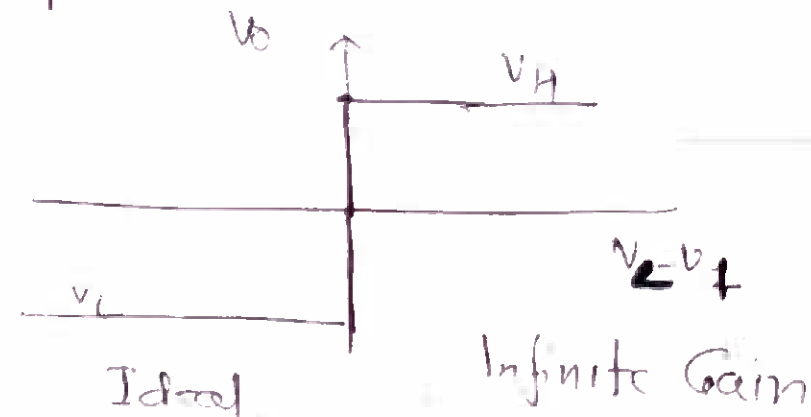


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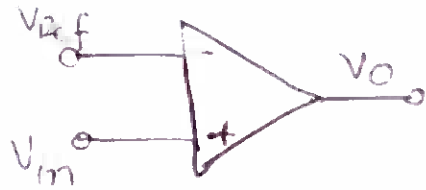


$$\text{Gain} = \frac{dV_o}{dV_{in}} = \frac{V_H - V_L}{2\delta}$$

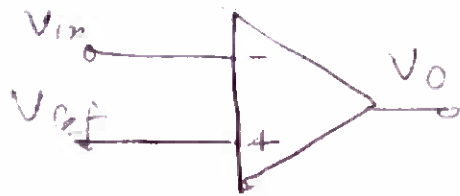
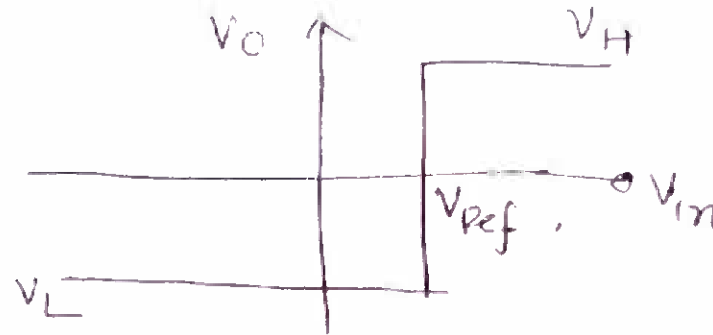
$$2\delta = \frac{10V}{10^5} = 10^{-4} = 0.1 \text{ mV} \\ = 100 \mu\text{V}$$



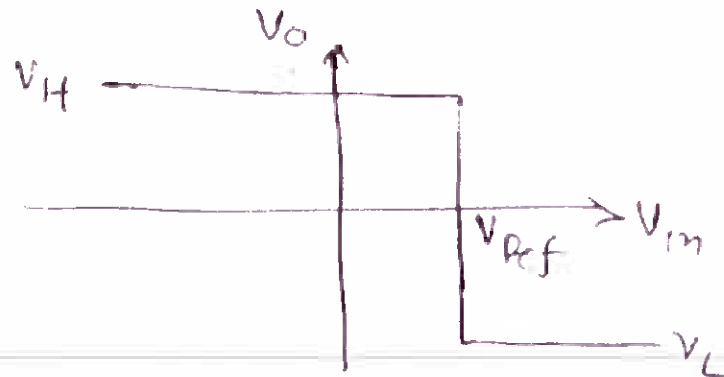
Slide No: 2



Non Inverting



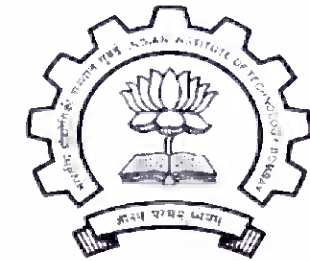
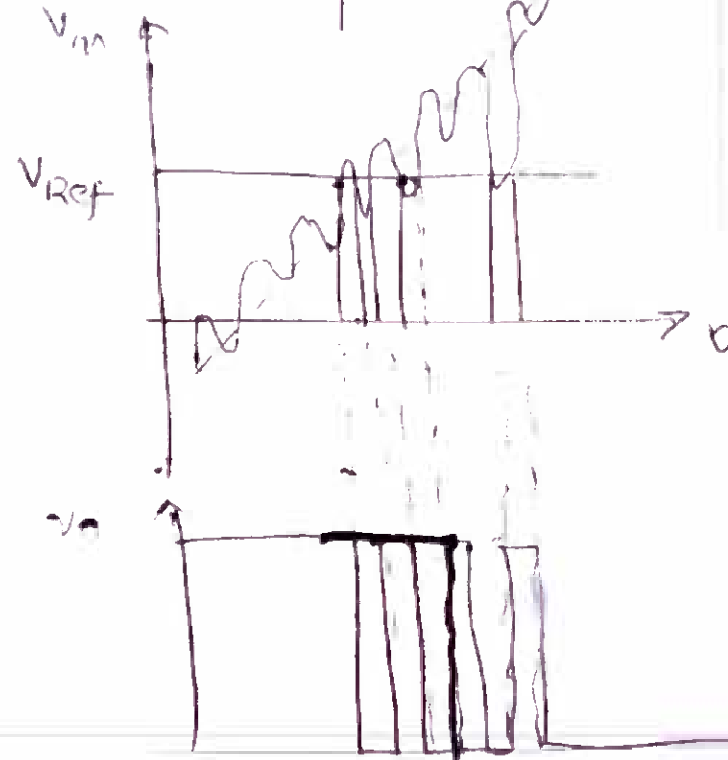
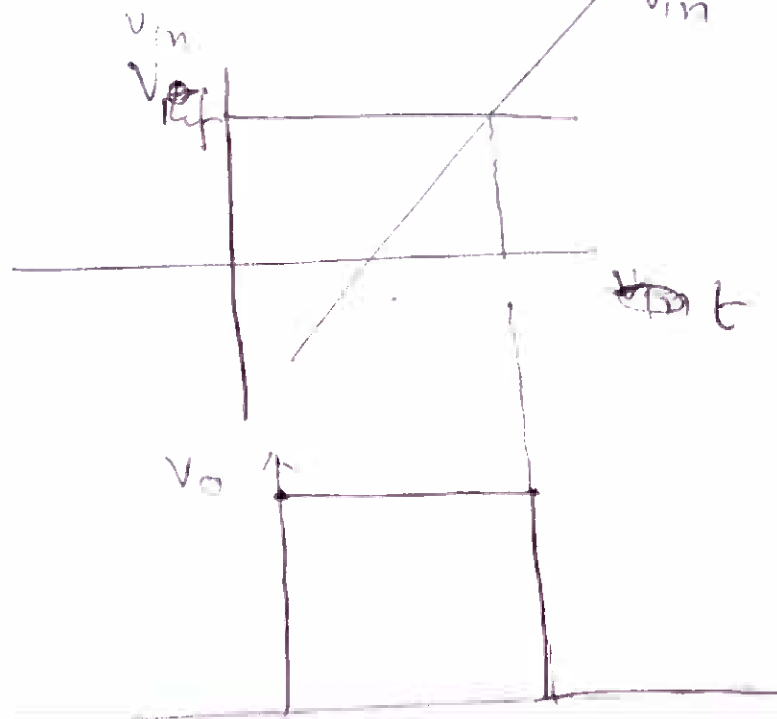
Inverting



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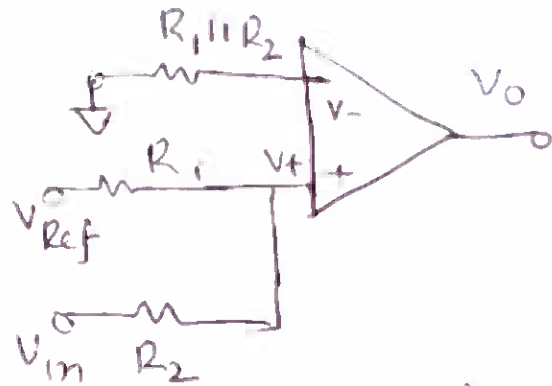
Slide No: 3

Problem with normal Comparator



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Slide No: 4



V_o switches when V_+ crosses V_- (0V here) i.e. at $V_+ = 0$
Crossover exists.

We find relationship between V_+ & V_{in} & V_{ref} now

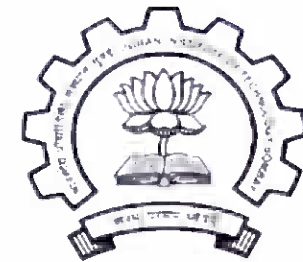
$$V_+ = \frac{R_2}{R_1 + R_2} V_{ref} + \frac{R_1}{R_1 + R_2} V_{in}$$

For Cross Over $V_+ \rightarrow 0$

$$\Rightarrow R_2 V_{ref} = R_1 V_{in} \quad \text{or} \quad V_{in}' = -\frac{R_2}{R_1} V_{ref} \text{ is Cross Over}$$

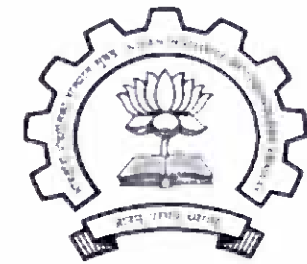
For Inverting case too cross over occurs at same value.

But V_o goes from High to Low



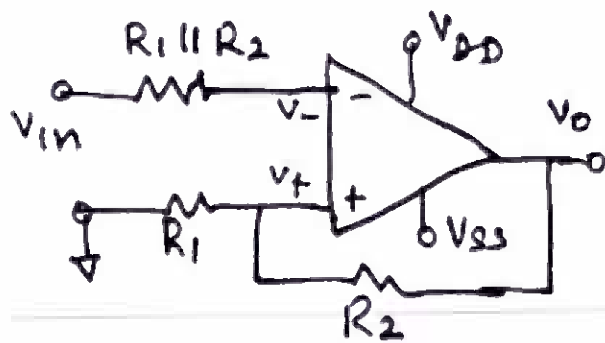
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Basic Schmitt Trigger



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Schmitt Trigger is modified Comparator with Positive Feedback. This results in a circuit which has TWO stable States (Bi Stable Multivibrator) of Oscillator



$$\text{Now } V_+ = \frac{R_1}{R_1 + R_2} \cdot V_o$$

Since $V_o = f(V_{in})$, hence V_+ is time-varying function

Depending upon whether $V_+ > V_-$
 $\vee V_- > V_+$, V_o can be +tive (V_{DD}) or
 -ve (V_{SS})

As $V_- = V_{in}$, V_o could be +tive or -ve depending upon V_{in} value & polarity.

Slide No: 6

We say $V_o = V_H$ (could be V_{DD})

or $V_o = V_L$ (could be V_{SS})

When $V_o = V_H$, obviously $v_- < v_+$

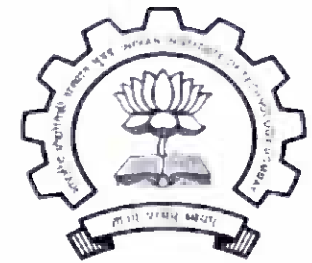
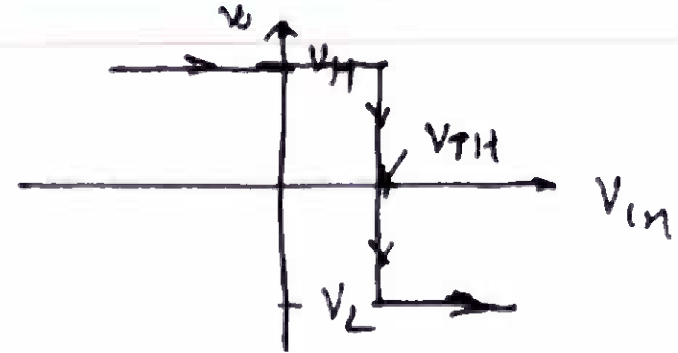
where $v_+ = \frac{R_1}{R_1 + R_2} V_H$

As long as $v_- < v_+$ $V_o = V_H$. (i.e. $V_{in} < V_+$ condition)

However if $V_{in} > V_+$ (or $v_- > v_+$), $V_o = V_L$

At $V_{in} = \frac{R_1}{R_1 + R_2} V_H$ cross over from $V_o = V_H$ to $V_o = V_L$ occurs

Then $V_{in} \Rightarrow$ called Cross-Over Voltage V_{TH} .

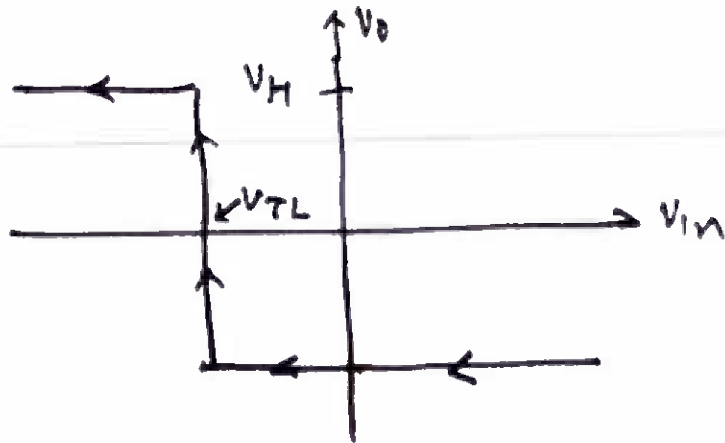


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Now let us say we have $V_0 = V_L$
(-ve) value. , then $V_- > V_+$ or $V_{in} > V_T$
or $V_{in} > V_{TL}$
 $\therefore V_T = \frac{R_1}{R_1 + R_2} \cdot V_L = V_{TL}$

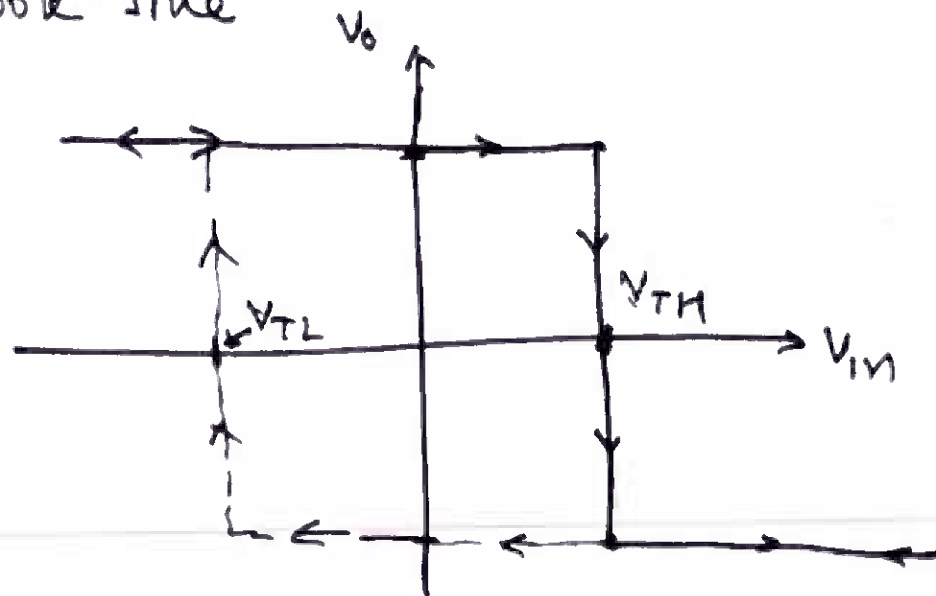
If we now reduce V_{in} (hence V_-), We see
that till $V_{in} > V_{TL}$, $V_0 = V_L$, but once $V_{in} \leq V_{TL}$
 V_0 switches due to comparator action to +ive V_H
and remains V_H till $V_{in} < V_{TL}$ ($V_- < V_+$).



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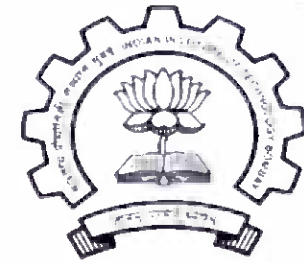
If V_{in} goes from -ve value ($< V_{TL}$) to +ve value ($> V_{TH}$) and returns back, the Transfer Characteristics

will look like



We thus see Hysteresis Effect.

Width of Hysteresis = $V_{TH} - V_{TL}$



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Clearly cross over points

V_{TH} & V_{TL} occur

when V_{in} crosses V_{TH}

and later $V_{in} < V_{TL}$

Example: If $V_{TH} = |V_{TL}| = 10V$

Evaluate R_1 for

$V_{TH} = -V_{TL} = 2V$

if $R_2 = 20K$

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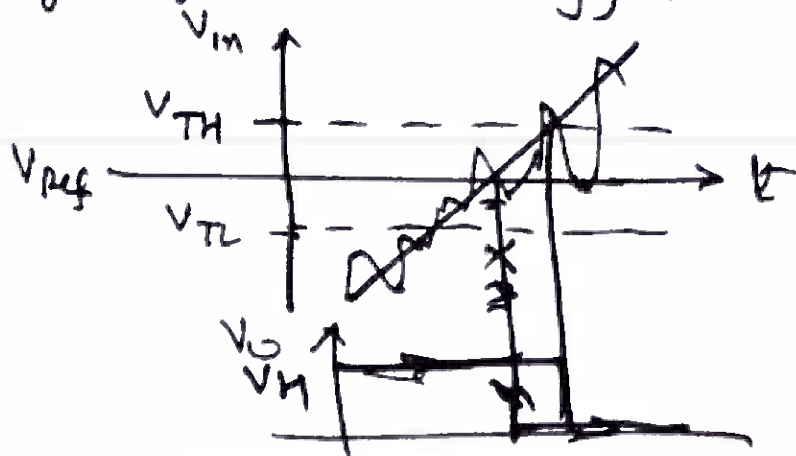
We have

$$V_{TH} = 2V = \frac{R_1}{R_1 + R_2} V_H = \frac{1}{\left(1 + \frac{20}{R_1}\right)} \cdot 10$$

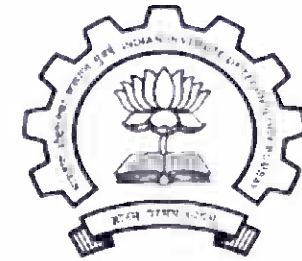
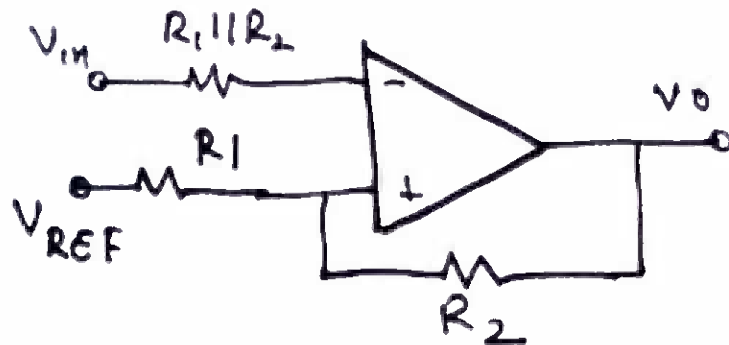
$$\begin{aligned} \therefore 2\left(1 + \frac{20}{R_1}\right) &= 10 & \text{or } 10R_1 &= 42 \\ \text{or } R_1 &= 4.2 \text{ K} \end{aligned}$$

similarly $V_{TL} = -2V = \frac{1}{1 + \frac{20}{R_1}} (-10)$, which gives $R_1 = 4.2 \text{ K}$

Advantages of Schmitt Trigger



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Schmitt Trigger with V_{REF} .CDEEP
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We define voltage

$$V_S = \frac{V_{TH} - V_{TL}}{2}$$

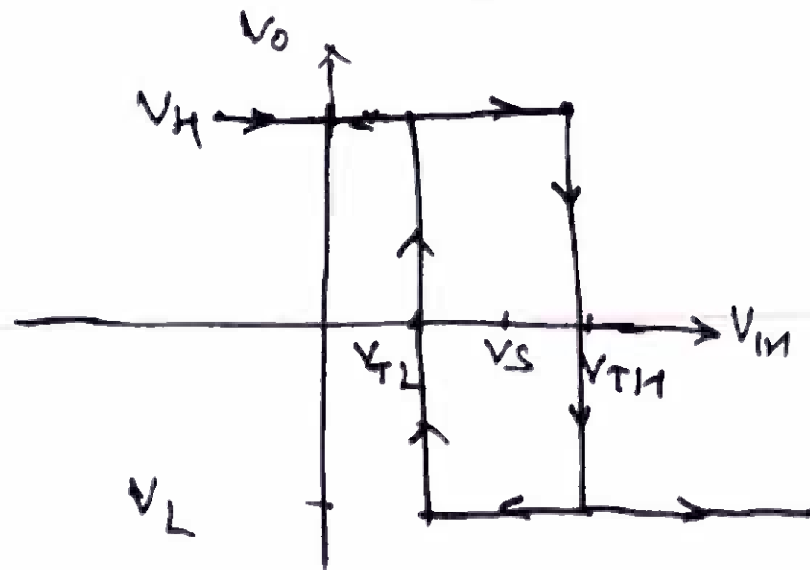
(Symmetric cross-over)

 V_S is called Switching VoltageIf $V_H = -V_L$ then

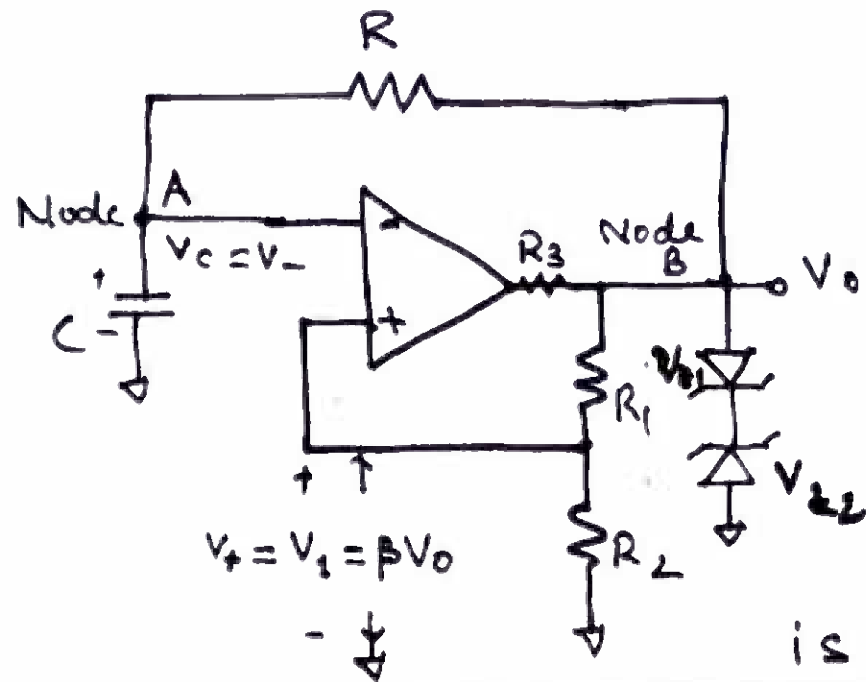
$$V_S = \frac{R_2}{R_1 + R_2} \cdot V_{REF}$$

$$V_{TH} = V_S + \frac{R_1}{R_1 + R_2} V_H$$

$$V_{TL} = V_S + \frac{R_1}{R_1 + R_2} V_L$$



Schmitt Trigger Oscillator

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Inverting Schmitt Trigger with 'limiters' can be used to create Free-Running Squarewave Generator, Also known as Astable Multivibrator

We know Capacitor Charging Transient is $V_C(t) = V_0 [1 - e^{-t/RC}]$ whose

RC is time constant

In Astable Multivibrator, whenever V_C crosses thresholds of comparator switching occurs leading to Square Wave Generation.

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$$\text{At } t = \frac{T}{2}, V_c(t) = +\beta V_0$$

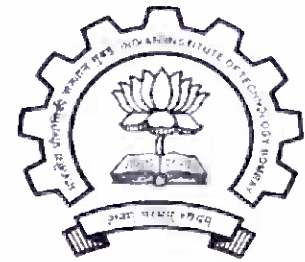
$$\text{Then } \beta V_0 = V_0 \left[1 - (1+\beta) e^{-T/2RC} \right]$$

$$= V_0 - (1+\beta) V_0 e^{-T/2RC}$$
$$\text{or } (1+\beta) e^{-T/2RC} = 1 - \beta$$

$$\text{or } T = 2RC \ln \left(\frac{1+\beta}{1-\beta} \right) = 2RC \ln \left[1 + \frac{2R_2}{R_1} \right]$$

Observe $T \neq f(V_0)$

Frequency Range 10 Hz to 10 kHz (why?)



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For v. high gain Comparator case we have

$$\begin{aligned} \text{(Then)} \quad V_o &= V_2 + V_D = V_o && \text{(Limiter limits } V_H \text{ \& } V_L \text{)} \\ \text{if } V_{id} &< 0 \\ & (V_- < V_+) \end{aligned}$$

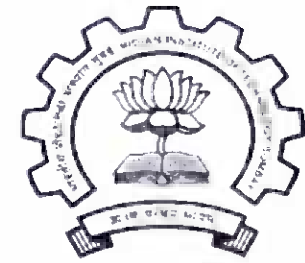
$$\Delta \quad V_o = -V_o \quad \text{if } V_{id} > 0$$

We start at an instant $t=0$, where $V_{id} < 0$ ($V_- < V_+$)

$$\text{Then } v_e(t) = V_o' = -\beta V_o \quad \text{(Initial condn)}$$

As $v_e(t)$ starts rising (capacitor starts charging) towards V_o with time constant $R \cdot C$, then

$$v_e(t) = V_o [1 - (1+\beta) e^{-t/RC}]$$

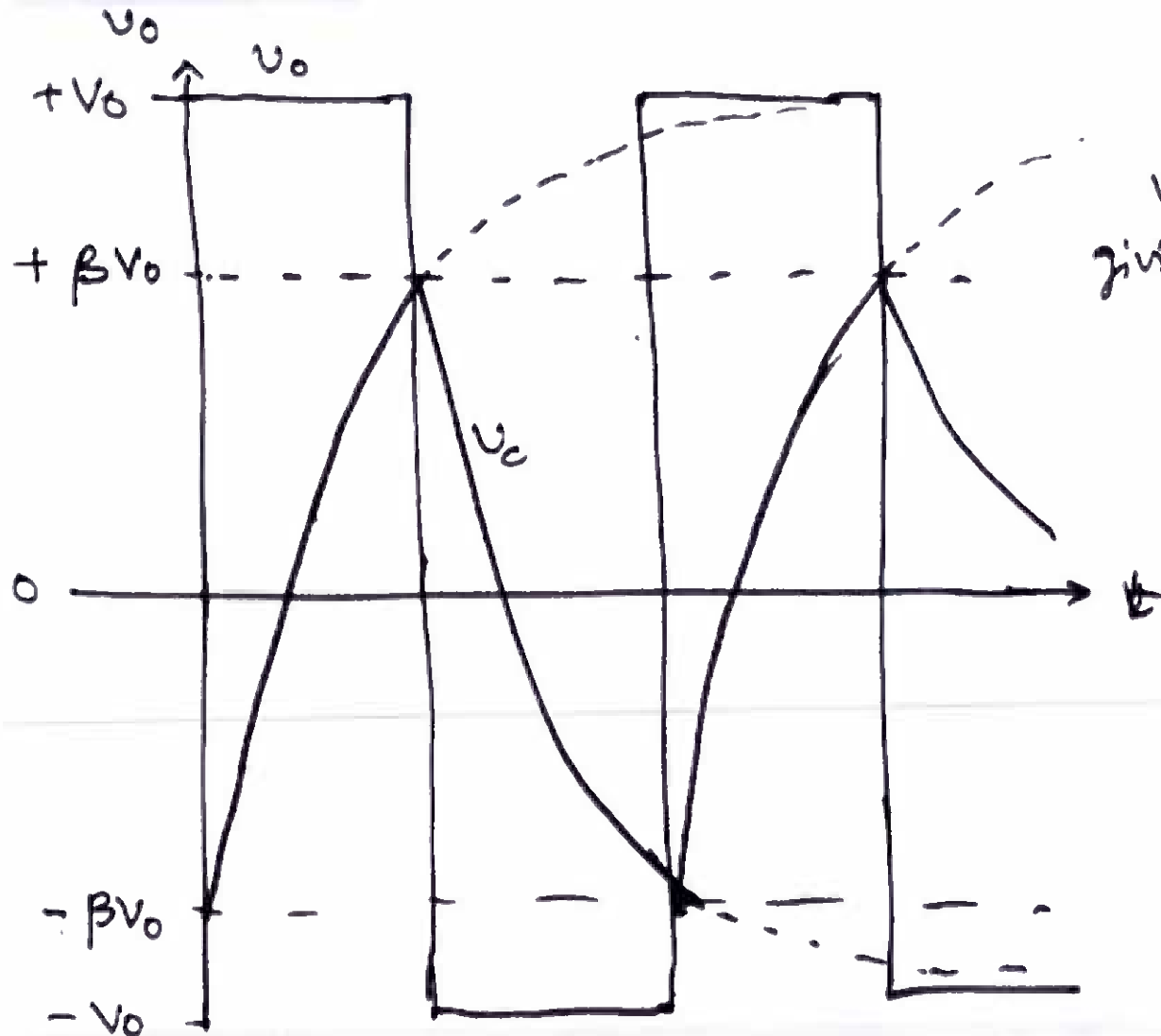


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Waveforms in SC Oscillator



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We have feedback
giving as $\frac{R_2 \cdot V_o}{R_1 + R_2} = \beta V_o$

Then Difference signal input is

$$V_{id} = V_- - V_+$$

$$= V_c - V_+$$

$$\text{But } V_+ = \frac{R_2}{R_1 + R_2} V_o$$

$$\therefore V_{id} = V_c - \frac{R_2}{R_1 + R_2} V_o$$

$$\text{or } V_{id} = V_c - \beta V_o$$