

Butterworth Filter

If $H(s)$ is Transfer F^n $V_o(s)/V_{in}(s)$
 given by $H(s) = \frac{A(s)}{B(s)}$

And if this has only Poles but 'no' zero $A(s) \text{ const.} = H_0$

$$\text{Then } H(s) = \frac{H_0}{B(s)}$$

Then the F^n $B^2(\omega) = 1 + \epsilon^2 \left(\frac{\omega}{\omega_0} \right)^{2n}$ is called

Butterworth Polynomial

$\epsilon \rightarrow 1$

The Filters using Butterworth F^n are called
 'Maximally Flat' kind, i.e. Ripple is v. low.



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$$|H(j\omega)| = \frac{H_0}{\sqrt{1 + \epsilon^2 \left(\frac{\omega}{\omega_0}\right)^{2N}}}$$

At $\omega = \omega_0$, $H(j\omega_0) = \frac{H_0}{\sqrt{1 + \epsilon^2}}$

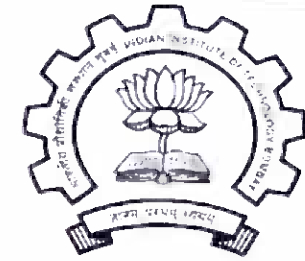
ϵ gives measure of Maximum Transmission A_{max}

$$A_{max} = 20 \log (1 + \epsilon^2)^{1/2}$$

If $\epsilon = 1$ $\epsilon = \sqrt{10^{A/10} - 1}$ gives Max. Transmission

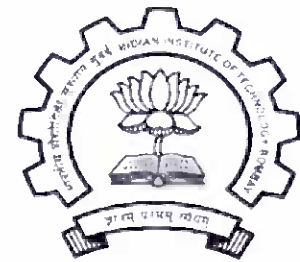
$$\left| \frac{H(j\omega)}{H_0} \right|^2 = \frac{1}{2} \quad \Delta \quad \underline{A_{min} = 0}$$

Passband terminates at $\omega = \omega_0$

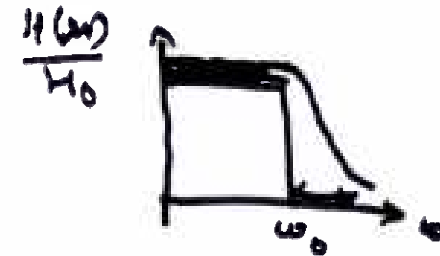
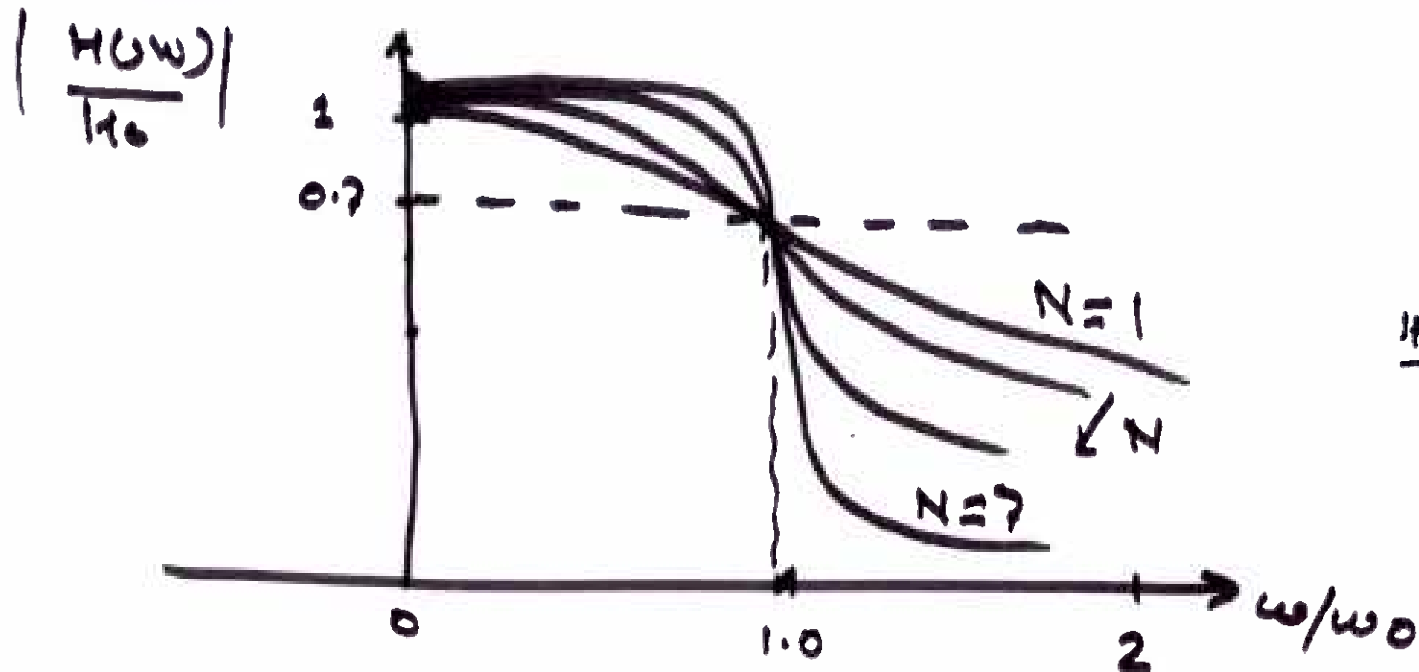


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Slide No 3



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At $w = w_0$

$$\left| \frac{H(jw)}{H_0} \right| = -20 \log \left[\frac{1}{\sqrt{1 + e^{2(\omega/\omega_0)^{2N}}}} \right]$$
$$= 10 \log \left[1 + e^{2(\frac{\omega}{\omega_0})^{2N}} \right]$$

\therefore Larger N means larger Attenuation

Slide No: 4

Example: For a LP Butterworth Filter, we need Attenuation of 40 db and at $\frac{\omega}{\omega_0} = 2$. We use $\epsilon = 1$.



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Then

$$\left| \frac{H(j\omega)}{H_0} \right|^2 = \frac{1}{1 + (\omega/\omega_0)^{2N}}$$

Given $\frac{H(j\omega)}{H_0} = \frac{1}{10^2} = 0.01$

$$\therefore 10^{-4} = \frac{1}{1 + 2^{2N}} \quad \text{or} \quad 2^{2N} = 10^4 - 1 \approx 10^4$$

$$\text{or} \quad 2N \log 2 = 4 \quad \text{or} \quad N = \frac{2}{\log 2} = \frac{2}{0.3010} \approx 6.64 \approx 7$$

Chebyshev Polynomial

$$C_N(\omega) = \cos\left(N \cos^{-1} \frac{\omega}{\omega_0}\right) \quad \omega \leq \omega_0$$

$$= \cosh\left(N \cosh^{-1} \left(\frac{\omega}{\omega_0}\right)\right) \quad \omega \geq \omega_0$$

Given $\left| \frac{H(j\omega)}{H_0} \right|^2 = (-40 \text{ dB})^2 = 10^{-4}$

$$\therefore 10^{-4} = \frac{1}{1 + (0.5089)^2 C_N^2(2)}$$

$$\text{or } C_N^2(2) = \frac{10^4 - 1}{(0.5089)^2} = \underline{3.86 \times 10^4}$$

$$C_N(2) = \sqrt{3.86 \times 10^4}$$

$$\text{or } 196.5 = \cosh\left(N \cosh^{-1}(2)\right)$$

Solving $N = 4.53 \approx 5$



Slide No: 6

The Ripple frequency ω_c is related to '-3db' cut-off frequency ω_H

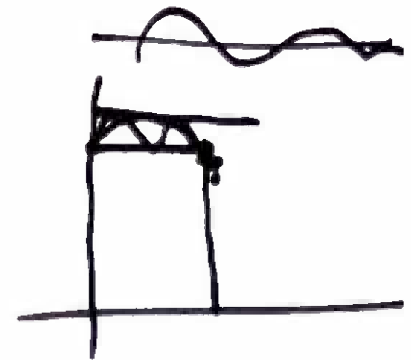
$$\omega_H = \omega_c \cosh \left(\frac{1}{N} \cosh^{-1} \frac{1}{\epsilon} \right)$$

For 1db ripple with say $N=5$ $\omega_H = 1.03 \omega_c$

Example: $\gamma = 1\text{db}$, Δ at $\frac{\omega}{\omega_0} = 2$ \therefore Attenuation is 40db.

Since $\gamma = 1\text{db}$, $\epsilon = 0.5089$

$$\therefore \left| \frac{H(j\omega)}{H_0} \right|^2 = \frac{1}{1 + (0.5089)^2 C_n^2(2)}$$



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Slide No: 7

Chebyshev filters are 'All Pole' filters and has larger 'Ripple'

but sharper fall for lower number of sections (N) compared to Butterworth.

N represents number of Poles.

Parameter ϵ is related to passband ripple γ in db by

$$\epsilon^2 = 10^{\frac{(\gamma/10)}{2}} - 1$$

For 0.5 db ripple ($\gamma=0.5$), $\epsilon = 0.3493$

For 1.0 db ripple ($\gamma=1.0$), $\epsilon = 0.5089$



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The Chebyshev Filters

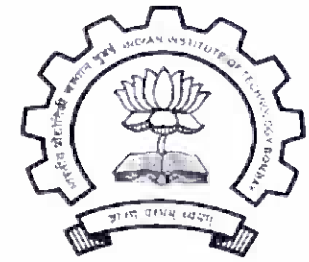
If the Transfer F^n has the form L of Magnitude as:

$$|H(j\omega)| = \frac{1}{\sqrt{1 + \epsilon^2 \cos^2 \left[N \cos^{-1} \left(\frac{\omega}{\omega_0} \right) \right]}} \quad \text{for } \omega \leq \omega_0$$

\longleftarrow $C_N(\cdot)$ \longrightarrow

$$= \frac{1}{\sqrt{1 + \epsilon^2 \cosh^2 \left[N \cosh^{-1} \left(\frac{\omega}{\omega_0} \right) \right]}} \quad \text{for } \omega \geq \omega_0$$

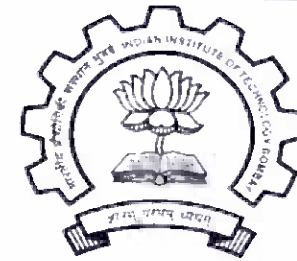
Then Transfer F^n represents Chebyshev Function



Slide No: 9

$$\text{and } \frac{\omega_H}{\omega_C} = \cosh\left(\frac{1}{5}\right) \cosh^{-1} \frac{1}{0.5089}$$
$$= 1.03$$

clearly we have a Sharper Fall at ω_H to ω_C
with $N=5$ and $\gamma = 1.0 \text{ dB}$

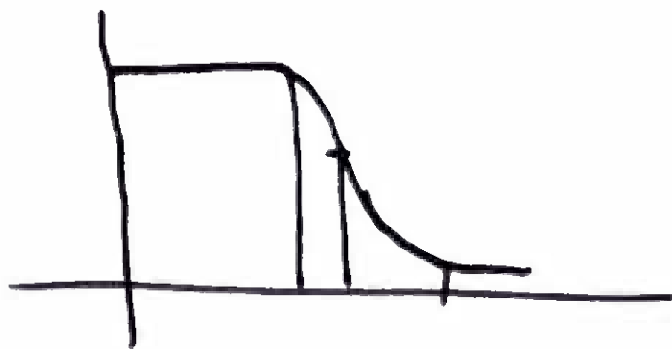


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Slide No: 10

Butterworth Polynomial

n	Factors of polynomial $B_n(s)$
1	$(s + 1)$
2	$(s^2 + 1.414s + 1)$
3	$(s + 1)(s^2 + s + 1)$
4	$(s^2 + 0.765s + 1)(s^2 + 1.848s + 1)$
5	$(s + 1)(s^2 + 0.618s + 1)(s^2 + 1.618s + 1)$
6	$(s^2 + 0.518s + 1)(s^2 + 1.414s + 1)(s^2 + 1.932s + 1)$
7	$(s + 1)(s^2 + 0.445s + 1)(s^2 + 1.247s + 1)(s^2 + 1.802s + 1)$
8	$(s^2 + 0.390s + 1)(s^2 + 1.111s + 1)(s^2 + 1.663s + 1)(s^2 + 1.962s + 1)$



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Chebyshev Polynomial

0.5-dB ripple ($\epsilon = 0.3493$)

1	$s + 2.863$
2	$s^2 + 1.425s + 1.516$
3	$(s + 0.626)(s^2 + 0.626s + 1.142)$
4	$(s^2 + 0.351s + 1.064)(s^2 + 0.845s + 0.356)$
5	$(s + 0.362)(s^2 + 0.224s + 1.036)(s^2 + 0.586s + 0.477)$
6	$(s^2 + 0.1554s + 1.024)(s^2 + 0.4142s + 0.5475)(s^2 + 0.5796s + 0.157)$
7	$(s + 0.2562)(s^2 + 0.1014s + 1.015)(s^2 + 0.3194s + 0.6657)(s^2 + 0.4616s + 0.2539)$
8	$(s^2 + 0.0872s + 1.012)(s^2 + 0.2484s + 0.7413)(s^2 + 0.3718s + 0.3872)(s^2 + 0.4386s + 0.08805)$

1.0-dB ripple ($\epsilon = 0.5089$)

1	$s + 1.965$
2	$(s^2 + 1.098s + 1.103)$
3	$(s + 0.494)(s^2 + 0.494s + 0.994)$
4	$(s^2 + 0.279s + 0.987)(s^2 + 0.674s + 0.279)$
5	$(s + 0.289)(s^2 + 0.179s + 0.988)(s^2 + 0.468s + 0.429)$
6	$(s^2 + 0.1244s + 0.9907)(s^2 + 0.3398s + 0.5577)(s^2 + 0.4642s + 0.1247)$
7	$(s + 0.2054)(s^2 + 0.0914s + 0.9927)(s^2 + 0.2562s + 0.6535)(s^2 + 0.3702s + 0.2304)$
8	$(s^2 + 0.07s + 0.9942)(s^2 + 0.1994s + 0.7236)(s^2 + 0.2994s + 0.3408)(s^2 + 0.3518s + 0.0702)$

Course Name
Analog Circuits

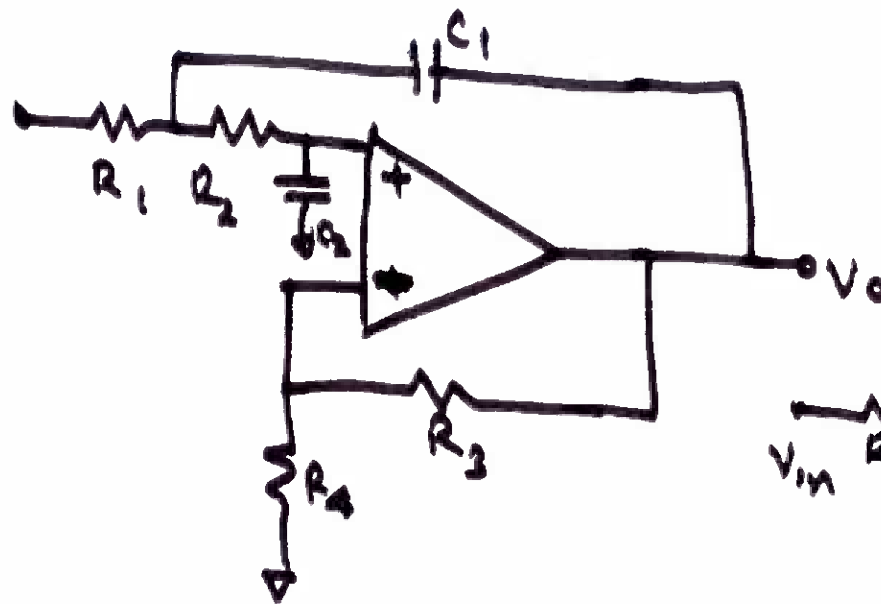
Lecture No. 22

Instructor's Name
Prof. A. N. Chandorkar

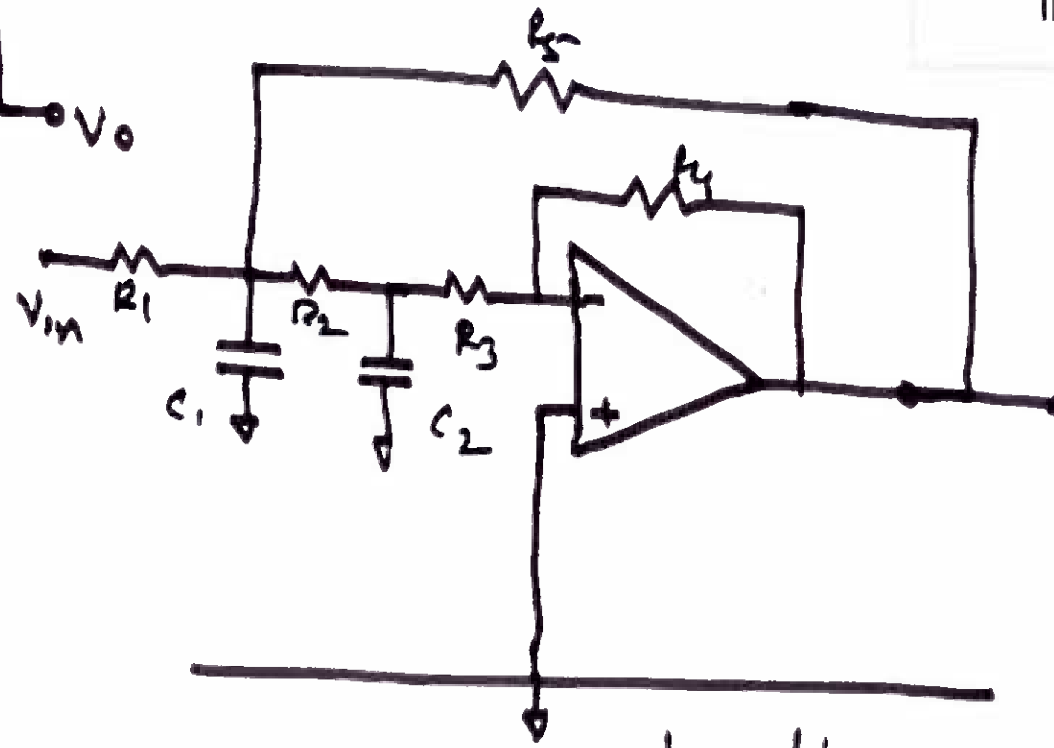
Sallen - Key Low Pass - Section



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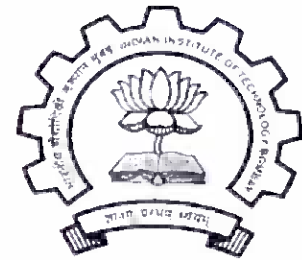


Non Inverting

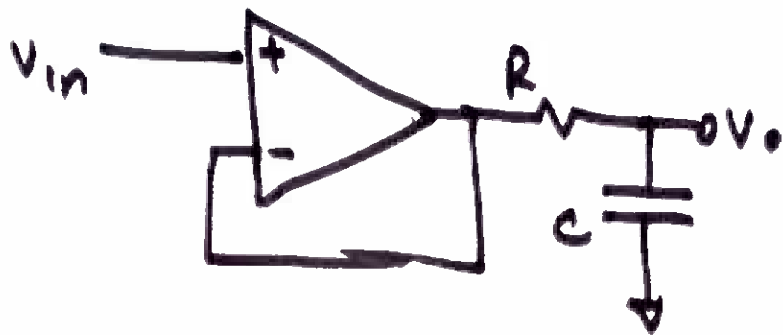


Inverting

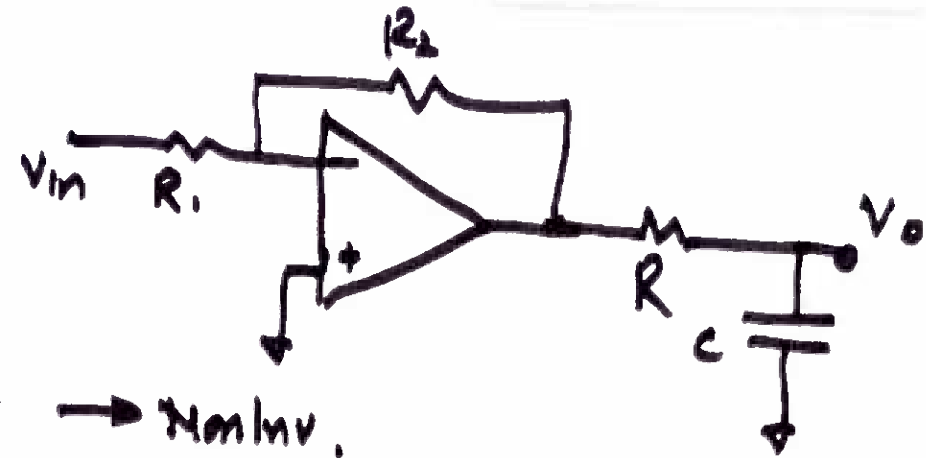
Creation of a Real Single Pole



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$$\frac{V_o}{V_{in}} = H(s) = \frac{1}{1 + RCs} = \frac{1}{1 + s/(C/R)}$$



→ NonInv.

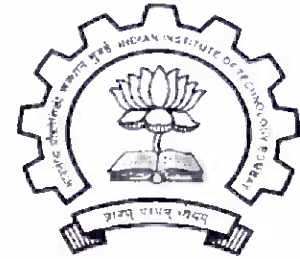
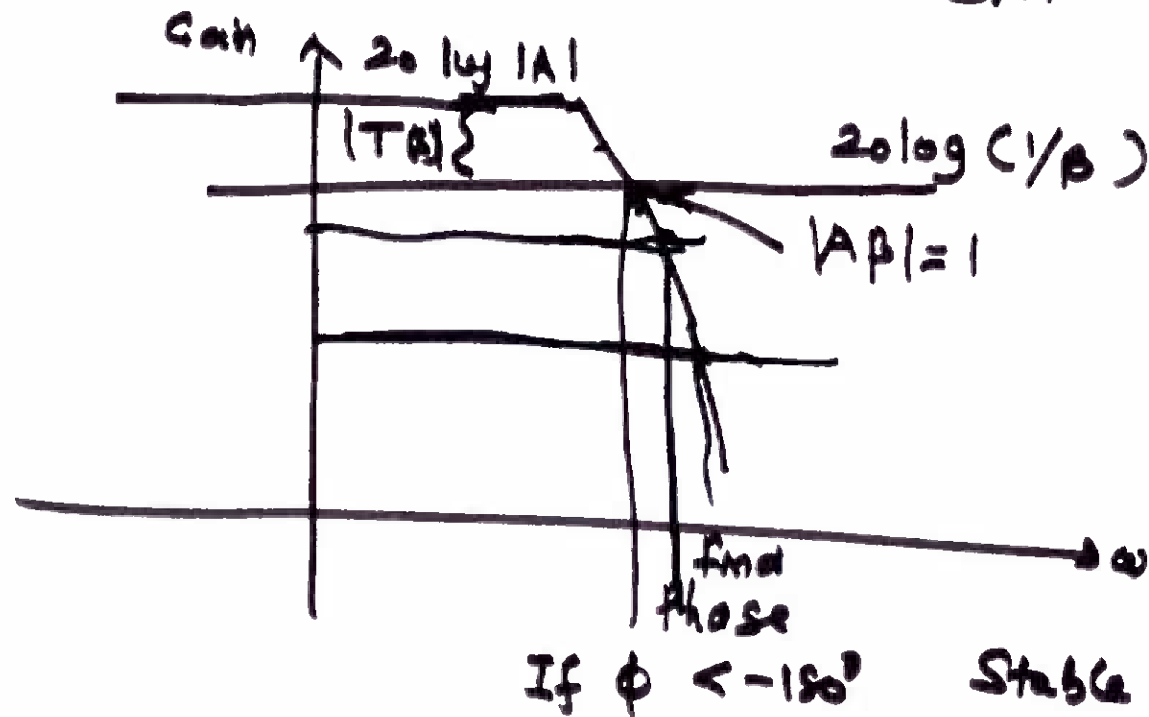
$$\frac{V_o}{V_{in}} = -\left(\frac{R_2}{R_1}\right) \cdot \frac{1}{1 + s/(C/R)}$$

Inverting

Stability : Revisit

We have $A\beta = \text{Loop Gain} = \frac{A}{1/\beta}$

$$\therefore 20 \log |A\beta| = 20 \log A - 20 \log (1/\beta)$$



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